

## DOUBLE CHANNEL HIGH SIDE SOLID STATE RELAY

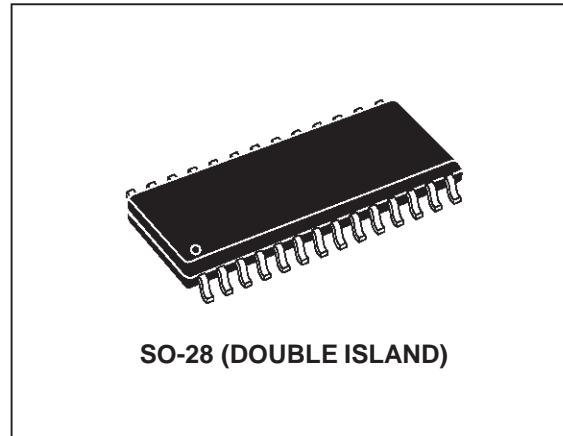
TYPE	$R_{DS(on)}$	$I_{OUT}$	$V_{CC}$
VND920	16mΩ	35 A (*)	36 V

(\*) Per channel with all the output pins connected to the PCB.

- CMOS COMPATIBLE INPUT
- PROPORTIONAL LOAD CURRENT SENSE
- SHORTED LOAD PROTECTION
- UNDervoltage AND OVERVOLTAGE SHUTDOWN
- OVERVOLTAGE CLAMP
- THERMAL SHUTDOWN
- CURRENT LIMITATION
- PROTECTION AGAINST LOSS OF GROUND AND LOSS OF  $V_{CC}$
- VERY LOW STAND-BY POWER DISSIPATION
- REVERSE BATTERY PROTECTION (\*)

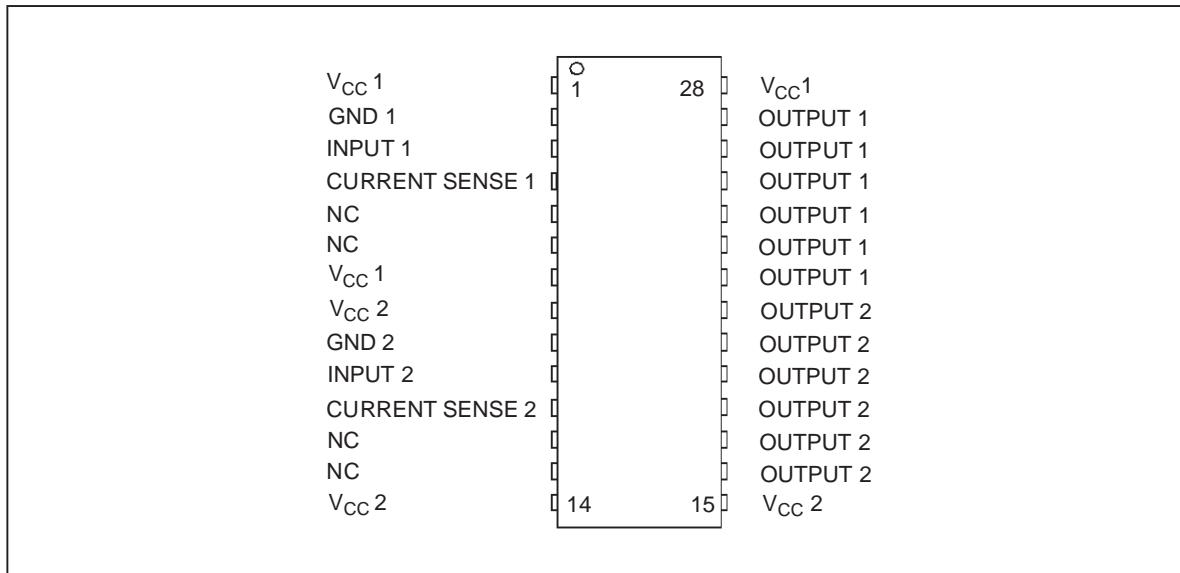
### DESCRIPTION

The VND920 is a double chip device made by using STMicroelectronics VIPower M0-3 Technology, intended for driving any kind of load with one side connected to ground. Active  $V_{CC}$  pin



voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table). Active current limitation combined with thermal shutdown and automatic restart protect the device against overload. Built-in analog current sense output delivers a current proportional to the load current. Device automatically turns off in case of ground pin disconnection.

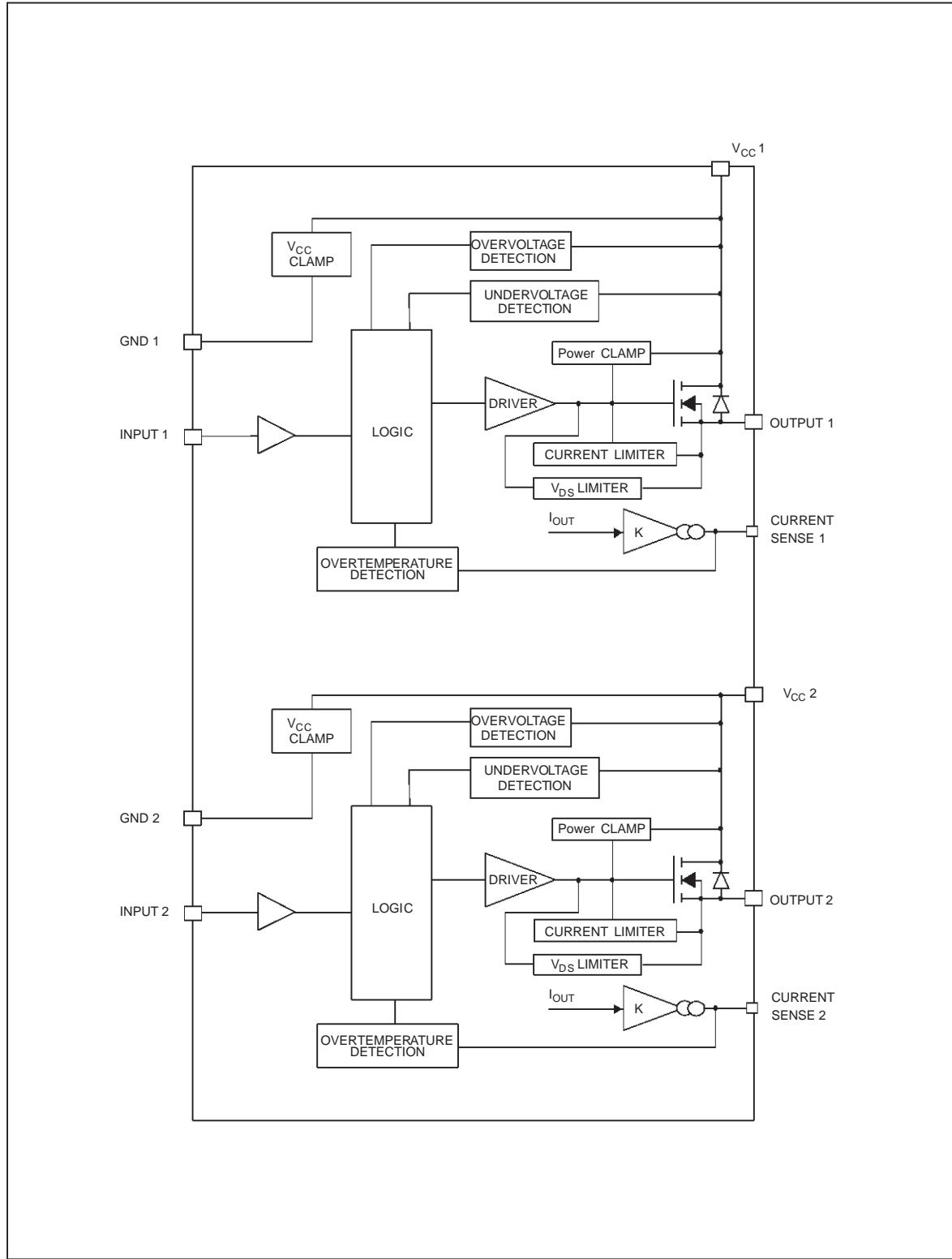
### CONNECTION DIAGRAM (TOP VIEW)



(\*) See application schematic at page 10

## VND920

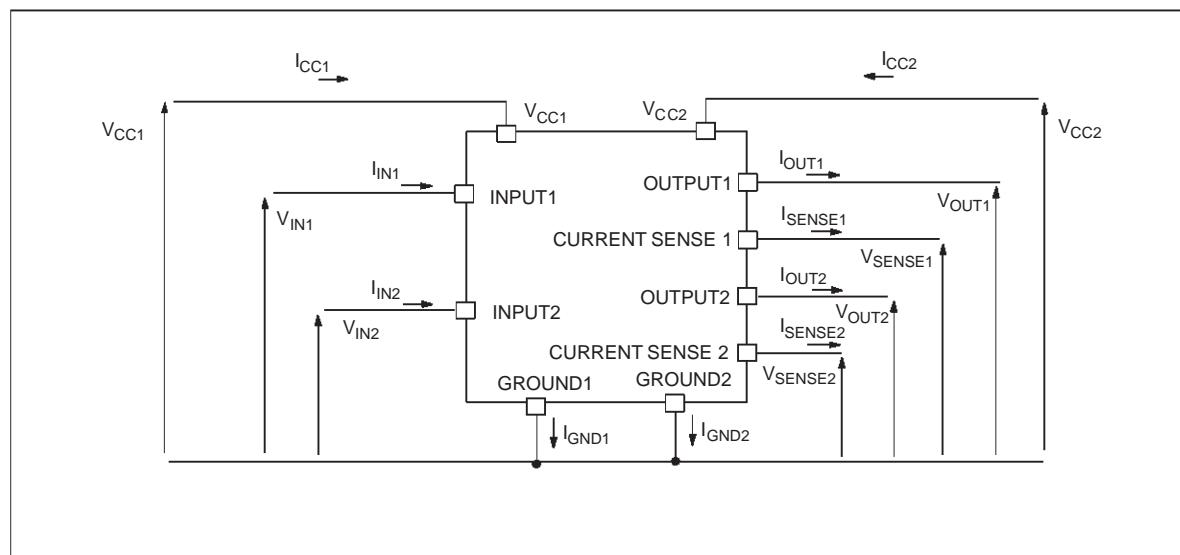
### BLOCK DIAGRAM



**ABSOLUTE MAXIMUM RATING (Per each channel)**

Symbol	Parameter	Value	Unit
$V_{CC}$	DC Supply Voltage	41	V
- $V_{CC}$	Reverse DC Supply Voltage	- 0.3	V
- $I_{GND}$	DC Reverse Ground Pin Current	- 200	mA
$I_{OUT}$	DC Output Current	Internally Limited	A
- $I_{OUT}$	Reverse DC Output Current	- 21	A
$I_{IN}$	DC Input Current	+/- 10	mA
$V_{CSENSE}$	Current Sense Maximum Voltage	-3 +15	V V
$V_{ESD}$	Electrostatic Discharge (Human Body Model: $R=1.5\text{ k}\Omega$ ; $C=100\text{ pF}$ ) - INPUT - CURRENT SENSE - OUTPUT - $V_{CC}$	4000 2000 5000 5000	V V V V
$E_{MAX}$	Maximum Switching Energy ( $L=0.25\text{ mH}$ ; $R_L=0\Omega$ ; $V_{bat}=13.5\text{ V}$ ; $T_{jstart}=150^\circ\text{C}$ ; $I_L=45\text{ A}$ )	350	mJ
$P_{tot}$	Power Dissipation $T \leq 25^\circ\text{C}$	6.25 (**)	W
$T_j$	Junction Operating Temperature	Internally limited	°C
$T_c$	Case Operating Temperature	- 40 to 150	°C
$T_{STG}$	Storage Temperature	- 55 to 150	°C

(\*\*) Per island

**CURRENT AND VOLTAGE CONVENTIONS**

## VND920

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### THERMAL DATA (Per island)

Symbol	Parameter	Value	Unit
$R_{thj\text{-lead}}$	Thermal Resistance Junction-lead	20	°C/W
$R_{thj\text{-amb}}$	Thermal Resistance Junction-ambient (one chip ON)	55 (*)	°C/W
$R_{thj\text{-amb}}$	Thermal Resistance Junction-ambient (two chips ON)	42 (*)	°C/W

(\*) When mounted on a standard single-sided FR-4 board with 100mm<sup>2</sup> of Cu (at least 35µm thick) connected to all V<sub>CC</sub> pins.  
Horizontal mounting and no artificial air flow.

### ELECTRICAL CHARACTERISTICS (8V < V<sub>CC</sub> < 36V; -40°C < T<sub>j</sub> < 150°C unless otherwise specified)

(Per island)

#### POWER

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V <sub>CC</sub>	Operating Supply Voltage		5.5	13	36	V
V <sub>USD</sub>	Undervoltage Shut-down		3	4	5.5	V
V <sub>OV</sub>	Overvoltage Shut-down		36			V
R <sub>ON</sub>	On State Resistance	I <sub>OUT</sub> =10A; T <sub>j</sub> =25°C I <sub>OUT</sub> =10A I <sub>OUT</sub> =3A; V <sub>CC</sub> =6V			16 32 55	mΩ
V <sub>clamp</sub>	Clamp Voltage	I <sub>CC</sub> =20mA (See note 1)	41	48	55	V
I <sub>S</sub>	Supply Current	Off State; V <sub>CC</sub> =13V; V <sub>IN</sub> =V <sub>OUT</sub> =0V Off State; V <sub>CC</sub> =13V; V <sub>IN</sub> =V <sub>OUT</sub> =0V; T <sub>j</sub> =25°C On State; V <sub>CC</sub> =13V; V <sub>IN</sub> =5V; I <sub>OUT</sub> =0A; R <sub>SENSE</sub> =3.9KΩ		10 10	25	µA
I <sub>L(off1)</sub>	Off State Output Current	V <sub>IN</sub> =V <sub>OUT</sub> =0V	0		50	µA
I <sub>L(off2)</sub>	Off State Output Current	V <sub>IN</sub> =0V; V <sub>OUT</sub> =3.5V	-75		0	µA
I <sub>L(off3)</sub>	Off State Output Current	V <sub>IN</sub> =V <sub>OUT</sub> =0V; V <sub>CC</sub> =13V; T <sub>j</sub> =125°C			5	µA
I <sub>L(off4)</sub>	Off State Output Current	V <sub>IN</sub> =V <sub>OUT</sub> =0V; V <sub>CC</sub> =13V; T <sub>j</sub> =25°C			3	µA

#### SWITCHING (V<sub>CC</sub>=13V)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
t <sub>d(on)</sub>	Turn-on Delay Time	R <sub>L</sub> =1.3Ω (see figure 2)		50		µs
t <sub>d(off)</sub>	Turn-off Delay Time	R <sub>L</sub> =1.3Ω (see figure 2)		50		µs
dV <sub>OUT</sub> /dt <sub>(on)</sub>	Turn-on Voltage Slope	R <sub>L</sub> =1.3Ω (see figure 2)		See relative diagram		V/µs
dV <sub>OUT</sub> /dt <sub>(off)</sub>	Turn-off Voltage Slope	R <sub>L</sub> =1.3Ω (see figure 2)		See relative diagram		V/µs

#### LOGIC INPUT

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V <sub>IL</sub>	Input Low Level				1.25	V
I <sub>IL</sub>	Low Level Input Current	V <sub>IN</sub> =1.25V	1			µA
V <sub>IH</sub>	Input High Level		3.25			V
I <sub>IH</sub>	High Level Input Current	V <sub>IN</sub> =3.25V			10	µA
V <sub>I(hyst)</sub>	Input Hysteresis Voltage		0.5			V
V <sub>ICL</sub>	Input Clamp Voltage	I <sub>IN</sub> =1mA I <sub>IN</sub> =-1mA	6	6.8 -0.7	8	V

Note 1: V<sub>clamp</sub> and V<sub>OV</sub> are correlated. Typical difference is 5V.

**ELECTRICAL CHARACTERISTICS (continued)**CURRENT SENSE ( $9V \leq V_{CC} \leq 16V$ ) (See Fig.1)

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
$K_1$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=1A; V_{SENSE}=0.5V;$ $T_j=-40^{\circ}C...150^{\circ}C$	3300	4400	6000	
$dK_1/K_1$	Current Sense Ratio Drift	$I_{OUT}=1A; V_{SENSE}=0.5V;$ $T_j=-40^{\circ}C...+150^{\circ}C$	-10		+10	%
$K_2$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=10A; V_{SENSE}=4V; T_j=-40^{\circ}C$ $T_j=25^{\circ}C...150^{\circ}C$	4200 4400	4900 4900	6000 5750	
$dK_2/K_2$	Current Sense Ratio Drift	$I_{OUT}=10A; V_{SENSE}=4V;$ $T_j=-40^{\circ}C...+150^{\circ}C$	-8		+8	%
$K_3$	$I_{OUT}/I_{SENSE}$	$I_{OUT}=30A; V_{SENSE}=4V; T_j=-40^{\circ}C$ $T_j=25^{\circ}C...150^{\circ}C$	4200 4400	4900 4900	5500 5250	
$dK_3/K_3$	Current Sense Ratio Drift	$I_{OUT}=30A; V_{SENSE}=4V;$ $T_j=-40^{\circ}C...+150^{\circ}C$	-6		+6	%
$I_{SENSEO}$	Analog Sense Leakage Current	$V_{CC}=6...16V; I_{OUT}=0A; V_{SENSE}=0V;$ $T_j=-40^{\circ}C...+150^{\circ}C$	0		10	$\mu A$
$V_{SENSE}$	Max Analog Sense Output Voltage	$V_{CC}=5.5V; I_{OUT}=5A; R_{SENSE}=10K\Omega$ $V_{CC}>8V; I_{OUT}=10A; R_{SENSE}=10K\Omega$	2 4			V
$V_{SENSEH}$	Sense Voltage in Overtemperature conditions	$V_{CC}=13V; R_{SENSE}=3.9K\Omega$		5.5		V
$R_{SENSE}$	Intrinsic sense pin resistance	$V_{CC}=13V; T_j>T_{TSD};$ Output Open		400		$\Omega$
$t_{DSENSE}$	Current sense delay response	to 90% $I_{SENSE}$ (see note 2)			500	$\mu s$

**PROTECTIONS**

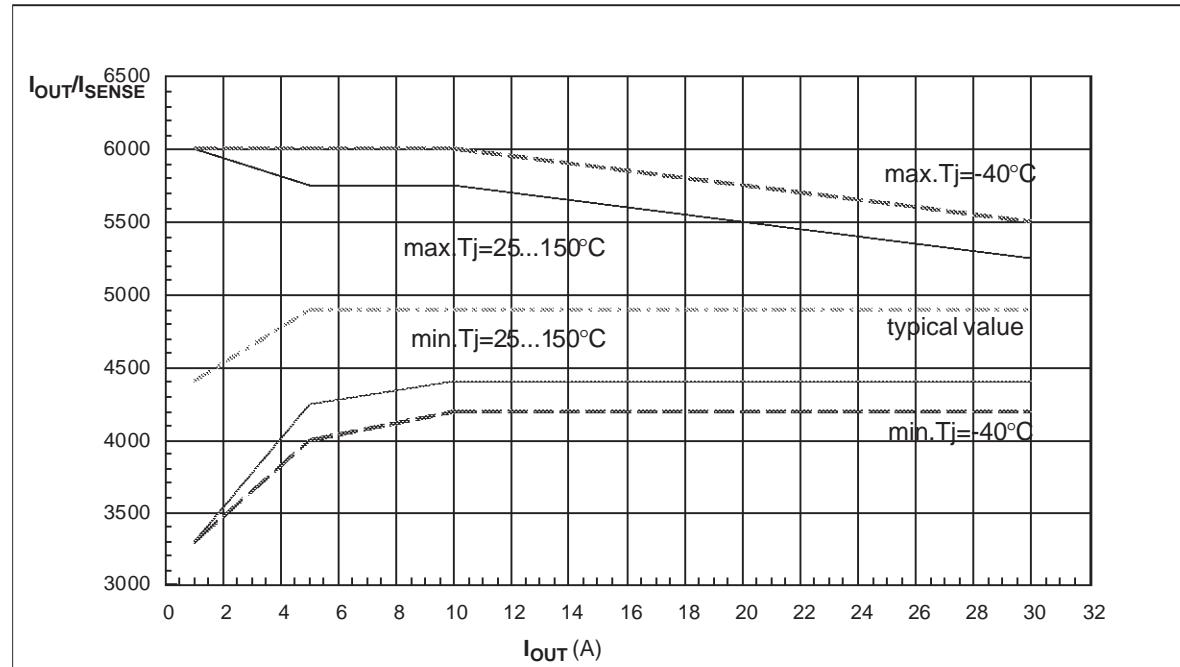
Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
$T_{TSD}$	Shut-down Temperature		150	175	200	$^{\circ}C$
$T_R$	Reset Temperature		135			$^{\circ}C$
$T_{hyst}$	Thermal Hysteresis		7	15		$^{\circ}C$
$I_{lim}$	DC Short Circuit Current	$V_{CC}=13V$ $5V < V_{CC} < 36V$	30	45	75	A
$V_{demag}$	Turn-off Output Clamp Voltage	$I_{OUT}=2A; V_{IN}=0V; L=6mH$	$V_{CC}-41$	$V_{CC}-48$	$V_{CC}-55$	V
$V_{ON}$	Output Voltage Drop Limitation	$I_{OUT}=1A; T_j=-40^{\circ}C...+150^{\circ}C$		50		mV

Note 2: current sense signal delay after positive input slope

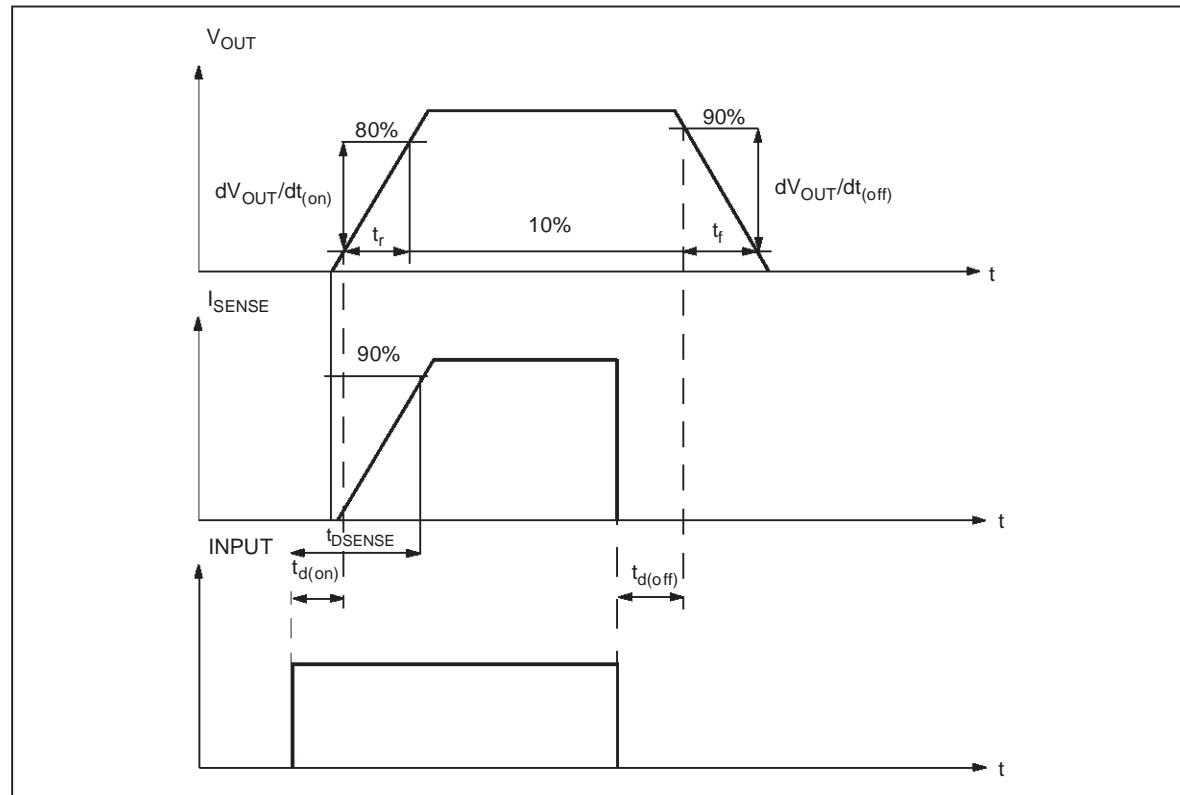
Note: Sense pin doesn't have to be left floating.

## VND920

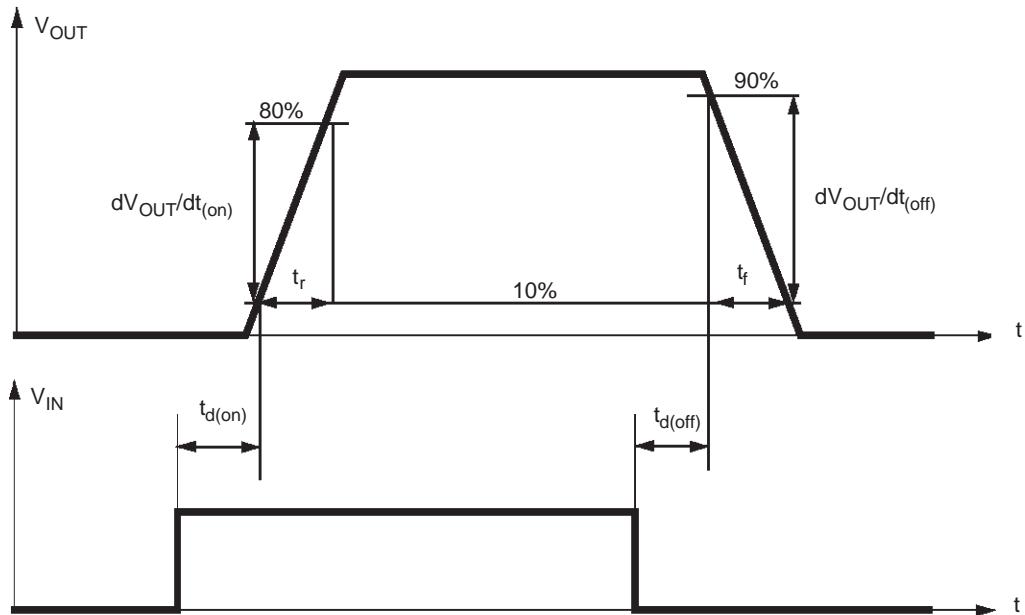
**Figure 1:**  $I_{OUT}/I_{SENSE}$  versus  $I_{OUT}$



**Figure 2:** Switching Characteristics (Resistive load  $R_L = 1.3\Omega$ )



### Switching time Waveforms



**TRUTH TABLE** (Per each channel)

CONDITIONS	INPUT	OUTPUT	CURRENT SENSE
Normal operation	L	L	0
	H	H	Nominal
Overtemperature	L	L	0
	H	L	$V_{SENSEH}$
Undervoltage	L	L	0
	H	L	0
Overvoltage	L	L	0
	H	L	0
Short circuit to GND	L	L	0
	H	L	$(T_j < T_{TSD}) 0$
	H	L	$(T_j > T_{TSD}) V_{SENSEH}$
Short circuit to $V_{CC}$	L	H	0
	H	H	< Nominal
Negative output voltage clamp	L	L	0

## VND920

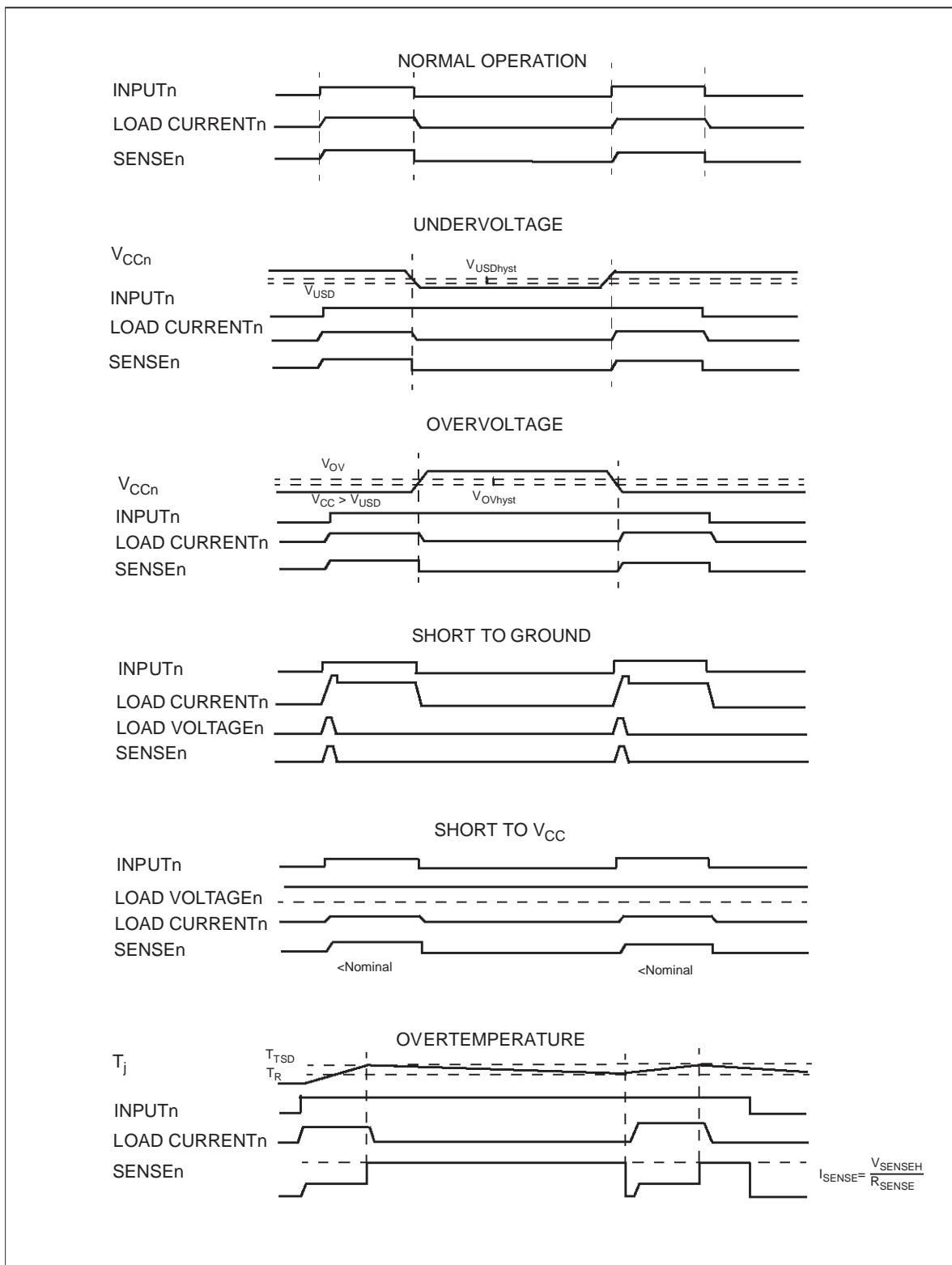
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### ELECTRICAL TRANSIENT REQUIREMENTS

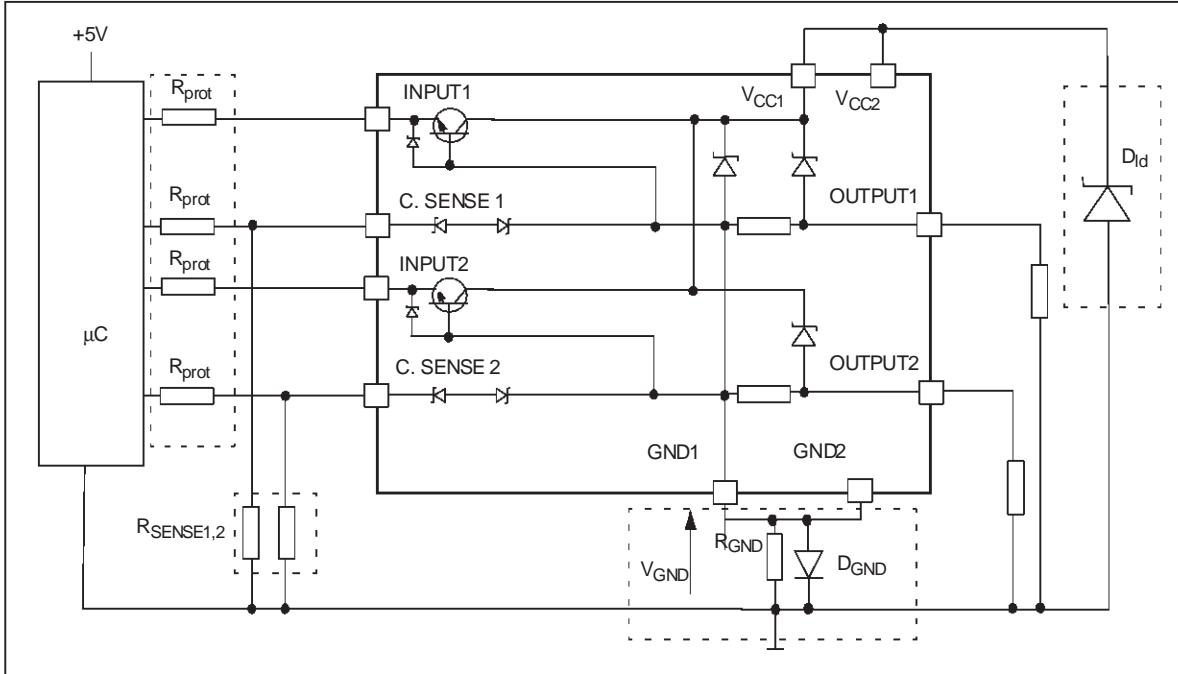
ISO T/R 7637/1 Test Pulse	TEST LEVELS				
	I	II	III	IV	Delays and Impedance
1	-25 V	-50 V	-75 V	-100 V	2 ms 10 Ω
2	+25 V	+50 V	+75 V	+100 V	0.2 ms 10 Ω
3a	-25 V	-50 V	-100 V	-150 V	0.1 μs 50 Ω
3b	+25 V	+50 V	+75 V	+100 V	0.1 μs 50 Ω
4	-4 V	-5 V	-6 V	-7 V	100 ms, 0.01 Ω
5	+26.5 V	+46.5 V	+66.5 V	+86.5 V	400 ms, 2 Ω

ISO T/R 7637/1 Test Pulse	TEST LEVELS RESULTS			
	I	II	III	IV
1	C	C	C	C
2	C	C	C	C
3a	C	C	C	C
3b	C	C	C	C
4	C	C	C	C
5	C	E	E	E

CLASS	CONTENTS
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device is not performed as designed after exposure to disturbance and cannot be returned to proper operation without replacing the device.

**Figure 3:** Waveforms

## APPLICATION SCHEMATIC



## GND PROTECTION NETWORK AGAINST REVERSE BATTERY

Solution 1: Resistor in the ground line ( $R_{GND}$  only). This can be used with any type of load.

The following is an indication on how to dimension the  $R_{GND}$  resistor.

- 1)  $R_{GND} \leq 600\text{mV} / (I_{S(on)\max})$ .
- 2)  $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where  $-I_{GND}$  is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device's datasheet.

Power Dissipation in  $R_{GND}$  (when  $V_{CC}<0$ : during reverse battery situations) is:

$$P_D = (-V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSD. Please note that the value of this resistor should be calculated with formula (1) where  $I_{S(on)\max}$  becomes the sum of the maximum on-state currents of the different devices.

Please note that if the microprocessor ground is not common with the device ground then the  $R_{GND}$  will produce a shift ( $I_{S(on)\max} * R_{GND}$ ) in the input thresholds and the status output values. This shift will vary depending on how many devices are ON in the case of several high side drivers sharing the same  $R_{GND}$ .

If the calculated power dissipation leads to a large resistor or several devices have to share the same resistor then the ST suggests to utilize Solution 2 (see below).

Solution 2: A diode ( $D_{GND}$ ) in the ground line.

A resistor ( $R_{GND}=1\text{k}\Omega$ ) should be inserted in parallel to  $D_{GND}$  if the device will be driving an inductive load.

This small signal diode can be safely shared amongst several different HSD. Also in this case, the presence of the ground network will produce a shift ( $\pm 600\text{mV}$ ) in the input threshold and the status output values if the microprocessor ground is not common with the device ground. This shift will not vary if more than one HSD shares the same diode/resistor network.

## LOAD DUMP PROTECTION

$D_{Id}$  is necessary (Transil or MOV) if the load dump peak voltage exceeds  $V_{CC}$  max DC rating. The same applies if the device will be subject to transients on the  $V_{CC}$  line that are greater than the ones shown in the ISO T/R 7637/1 table.

## μC I/Os PROTECTION:

If a ground protection network is used and negative transients are present on the  $V_{CC}$  line, the control pins will be pulled negative. ST suggests to insert a resistor ( $R_{prot}$ ) in line to prevent the μC I/Os pins to latch-up.

The value of these resistors is a compromise between the leakage current of μC and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of μC I/Os.

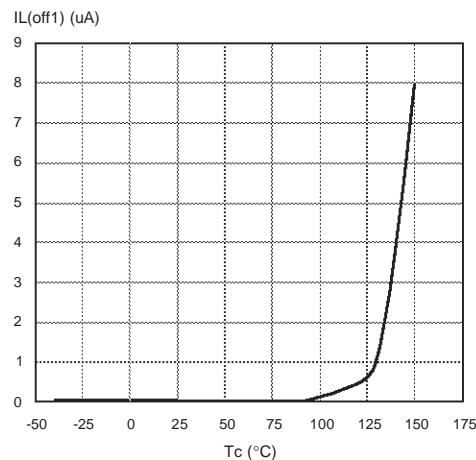
$$-V_{CCpeak} / I_{latchup} \leq R_{prot} \leq (V_{OH\mu C} - V_{IH} - V_{GND}) / I_{IHmax}$$

Calculation example:

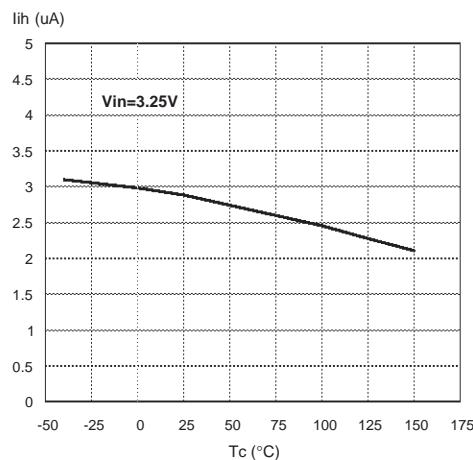
For  $V_{CCpeak} = -100\text{V}$  and  $I_{latchup} \geq 20\text{mA}$ ;  $V_{OH\mu C} \geq 4.5\text{V}$   
 $5\text{k}\Omega \leq R_{prot} \leq 65\text{k}\Omega$ .

Recommended  $R_{prot}$  value is  $10\text{k}\Omega$ .

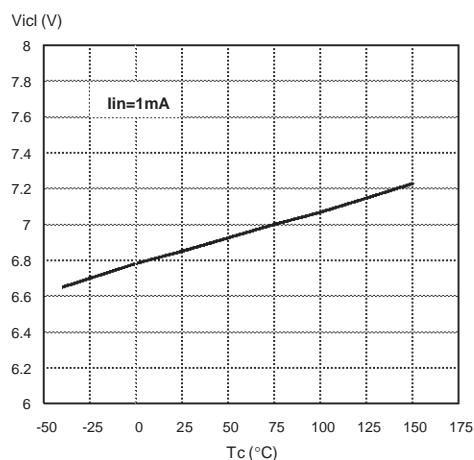
Off State Output Current



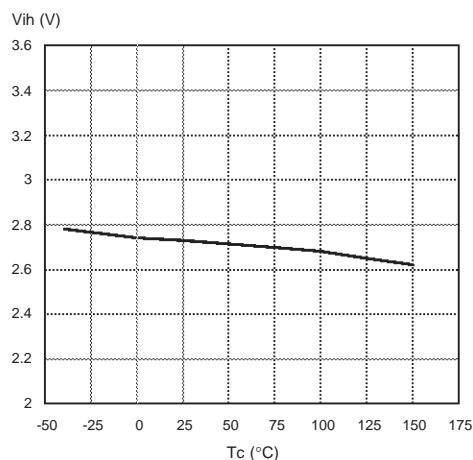
High Level Input Current



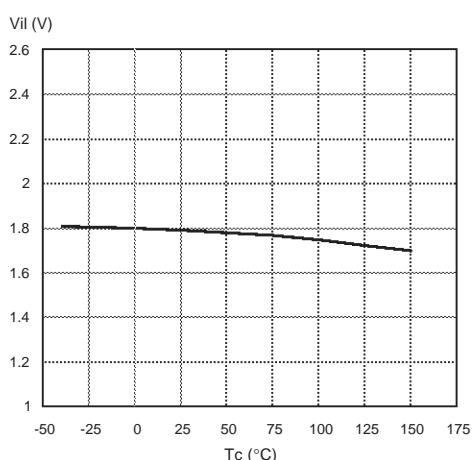
Input Clamp Voltage



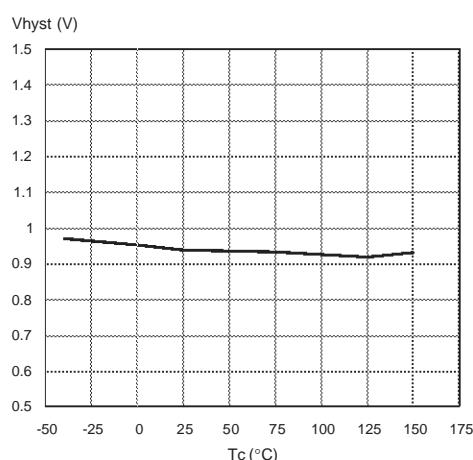
Input High Level



Input Low Level



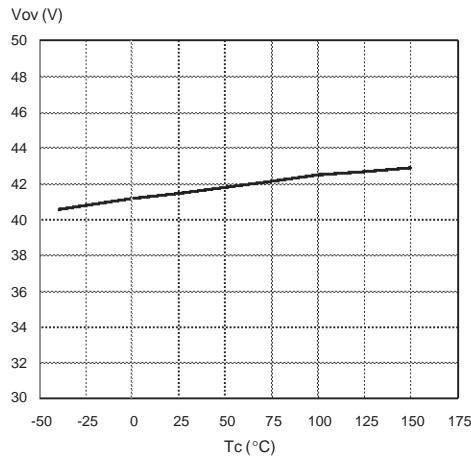
Input Hysteresis Voltage



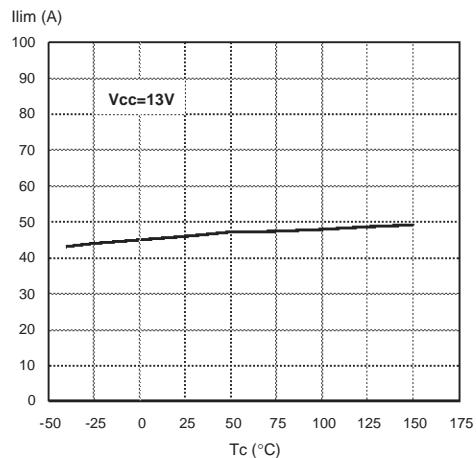
## VND920

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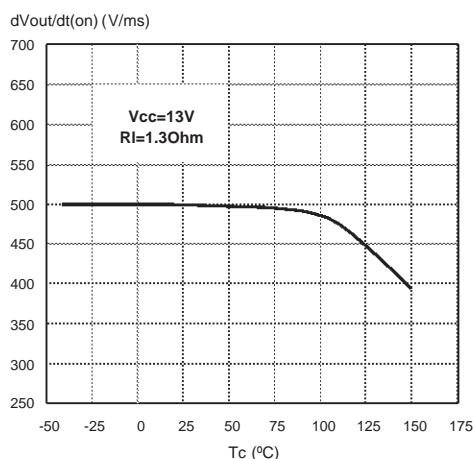
Overvoltage Shutdown



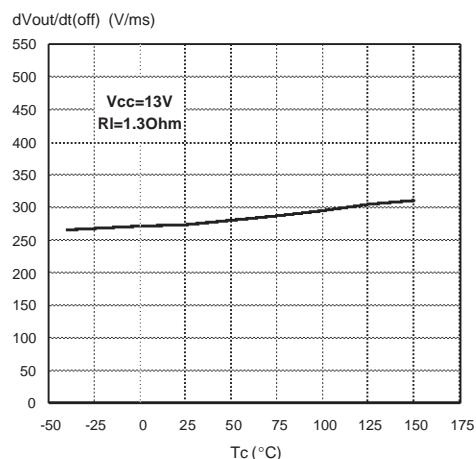
$I_{LIM}$  Vs  $T_{case}$



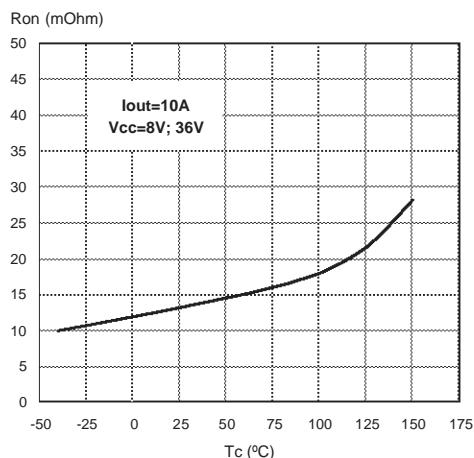
Turn-on Voltage Slope



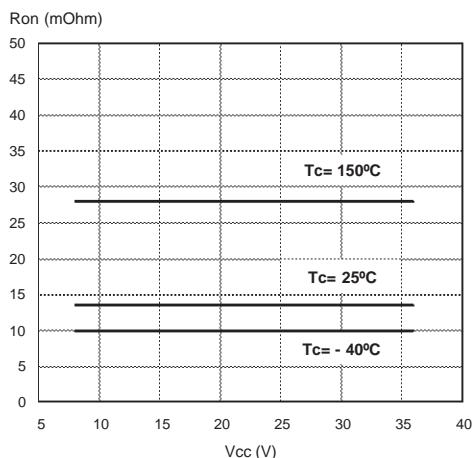
Turn-off Voltage Slope

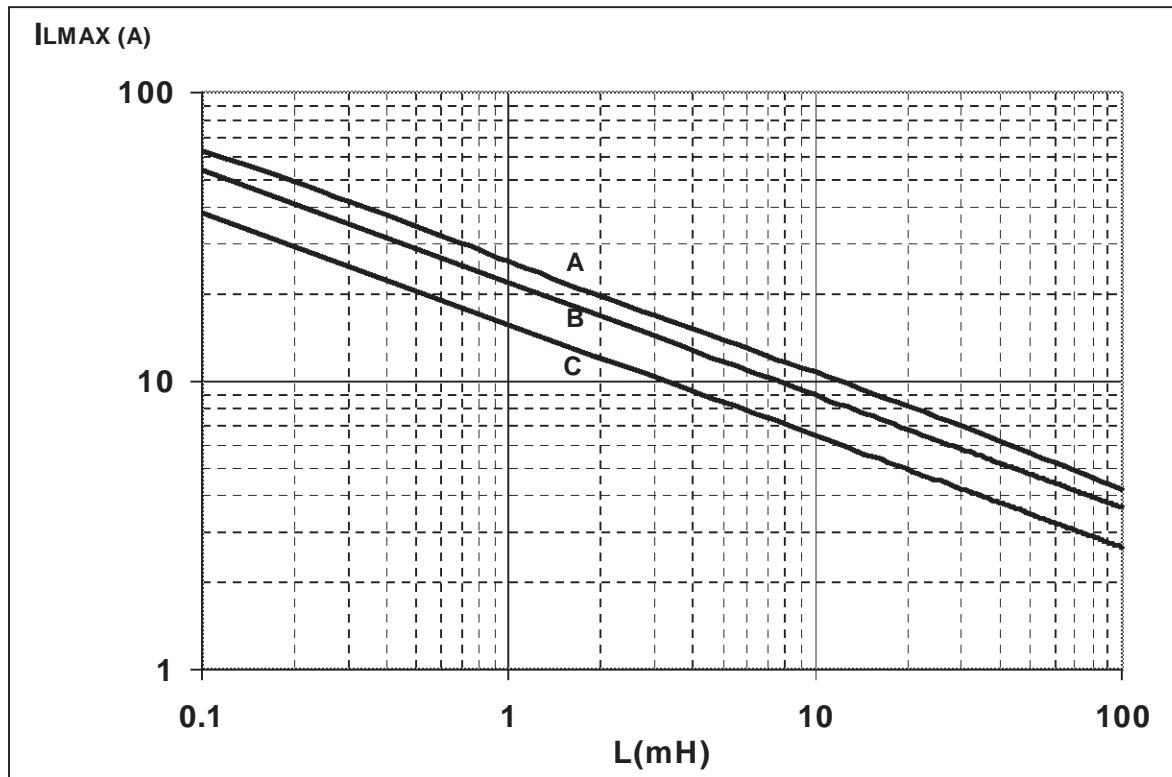
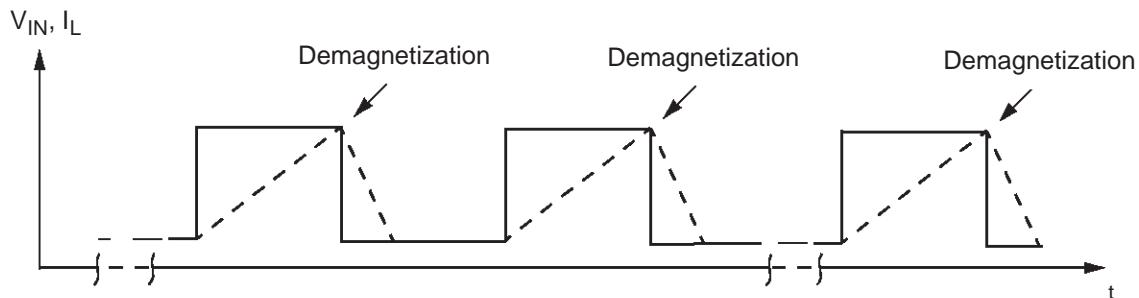


On State Resistance Vs  $T_{case}$



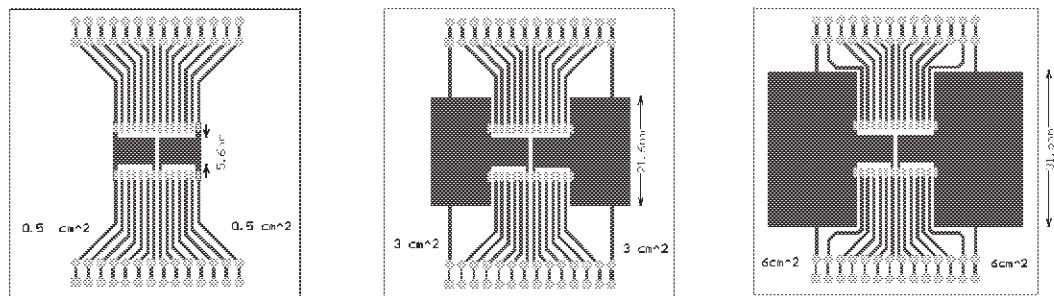
On State Resistance Vs  $V_{CC}$



**Maximum turn off current versus load inductance**A = Single Pulse at  $T_{j\text{start}}=150^\circ\text{C}$ B= Repetitive pulse at  $T_{j\text{start}}=100^\circ\text{C}$ C= Repetitive Pulse at  $T_{j\text{start}}=125^\circ\text{C}$ Conditions: $V_{CC}=13.5\text{V}$ Values are generated with  $R_L=0\Omega$ In case of repetitive pulses,  $T_{j\text{start}}$  (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves B and C.

## SO-28 DOUBLE ISLAND THERMAL DATA

### SO-28 Double island PC Board



Layout condition of  $R_{th}$  and  $Z_{th}$  measurements (PCB FR4 area= 58mm x 58mm, PCB thickness=2mm, Cu thickness=35 $\mu$ m, Copper areas: 0.5cm<sup>2</sup>, 3cm<sup>2</sup>, 6cm<sup>2</sup>).

### Thermal calculation according to the PCB heatsink area

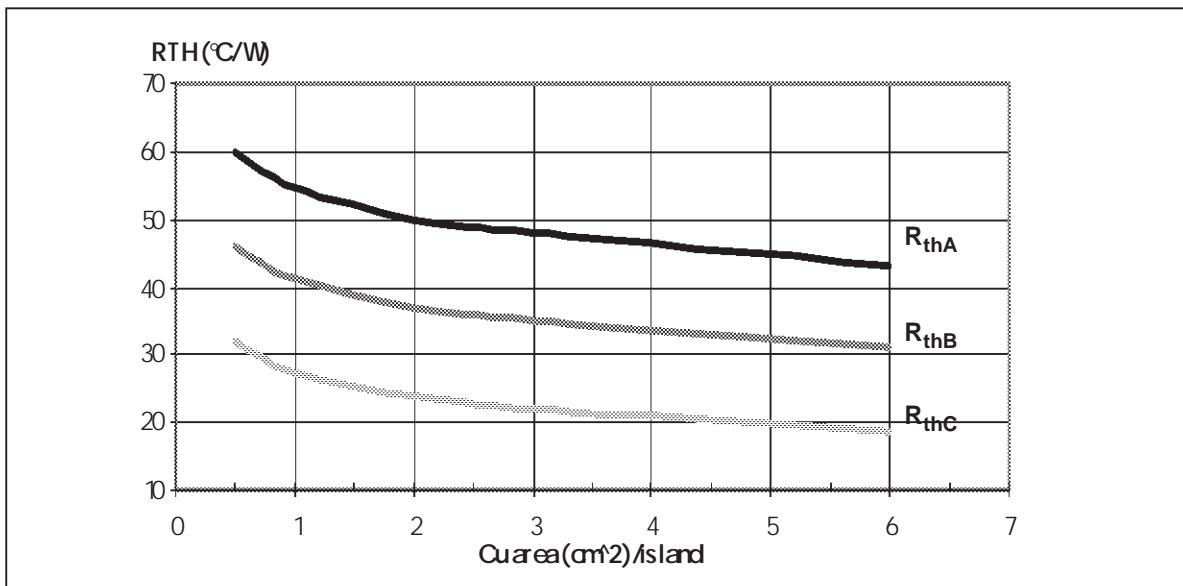
Chip 1	Chip 2	$T_{jchip1}$	$T_{jchip2}$	Note
ON	OFF	$R_{thA} \times P_{dchip1} + T_{amb}$	$R_{thC} \times P_{dchip1} + T_{amb}$	
OFF	ON	$R_{thC} \times P_{dchip2} + T_{amb}$	$R_{thA} \times P_{dchip2} + T_{amb}$	
ON	ON	$R_{thB} \times (P_{dchip1} + P_{dchip2}) + T_{amb}$	$R_{thB} \times (P_{dchip1} + P_{dchip2}) + T_{amb}$	$P_{dchip1}=P_{dchip2}$
ON	ON	$(R_{thA} \times P_{dchip1}) + R_{thC} \times P_{dchip2} + T_{amb}$	$(R_{thA} \times P_{dchip2}) + R_{thC} \times P_{dchip1} + T_{amb}$	$P_{dchip1} \neq P_{dchip2}$

$R_{thA}$  = Thermal resistance Junction to Ambient with one chip ON

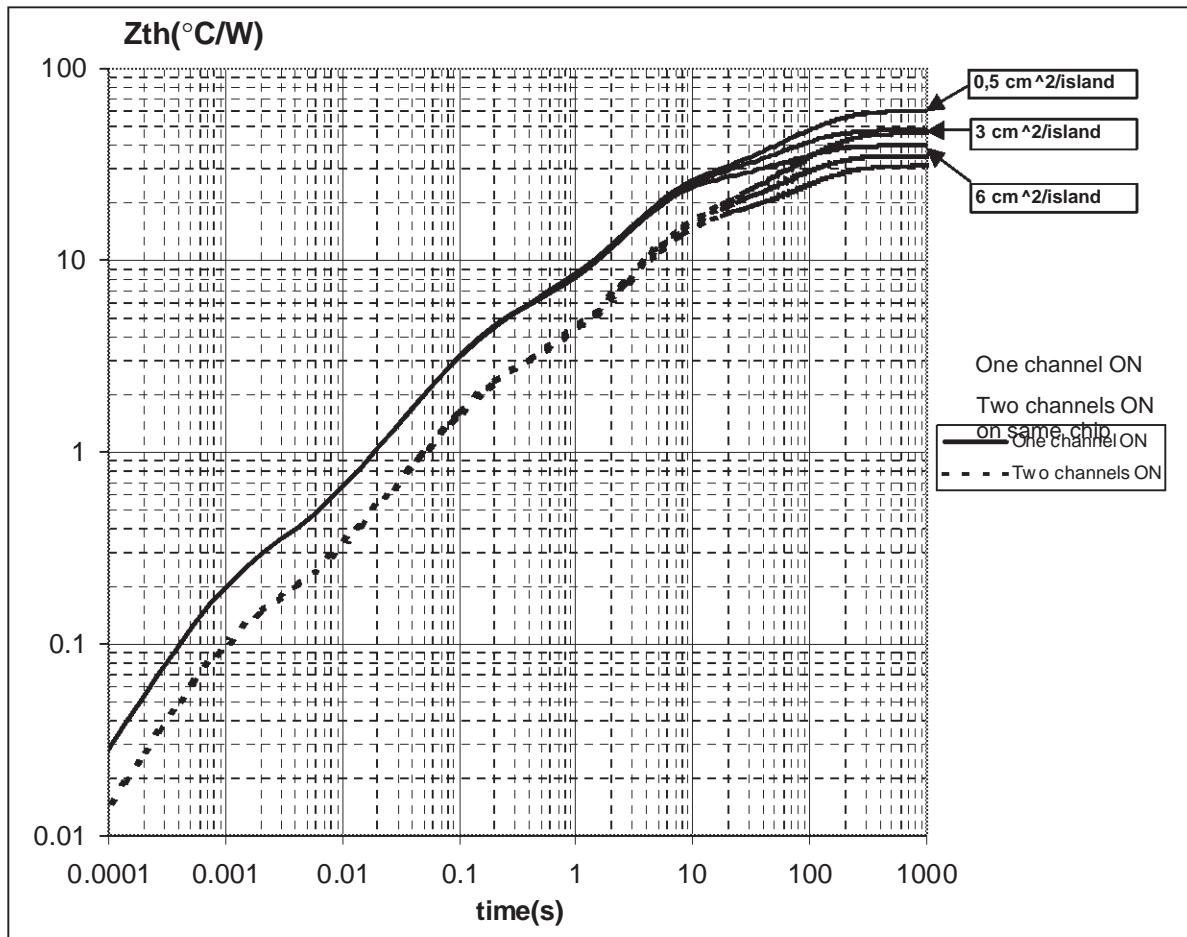
$R_{thB}$  = Thermal resistance Junction to Ambient with both chips ON and  $P_{dchip1}=P_{dchip2}$

$R_{thC}$  = Mutual thermal resistance

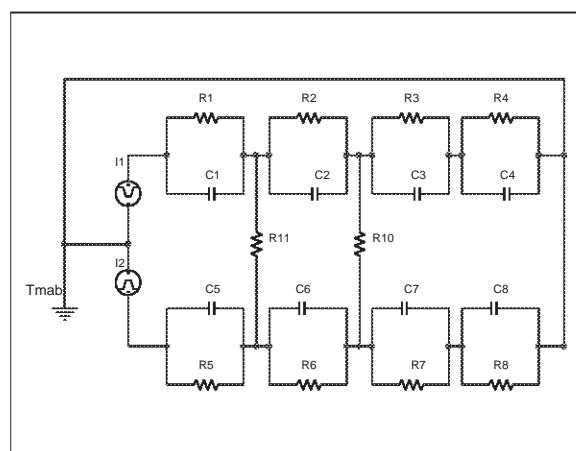
### $R_{thj-amb}$ Vs PCB copper area in open box free air condition



## Thermal Impedance Junction Ambient Single Pulse



Thermal fitting model of a two channels HSD in SO-28



Pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

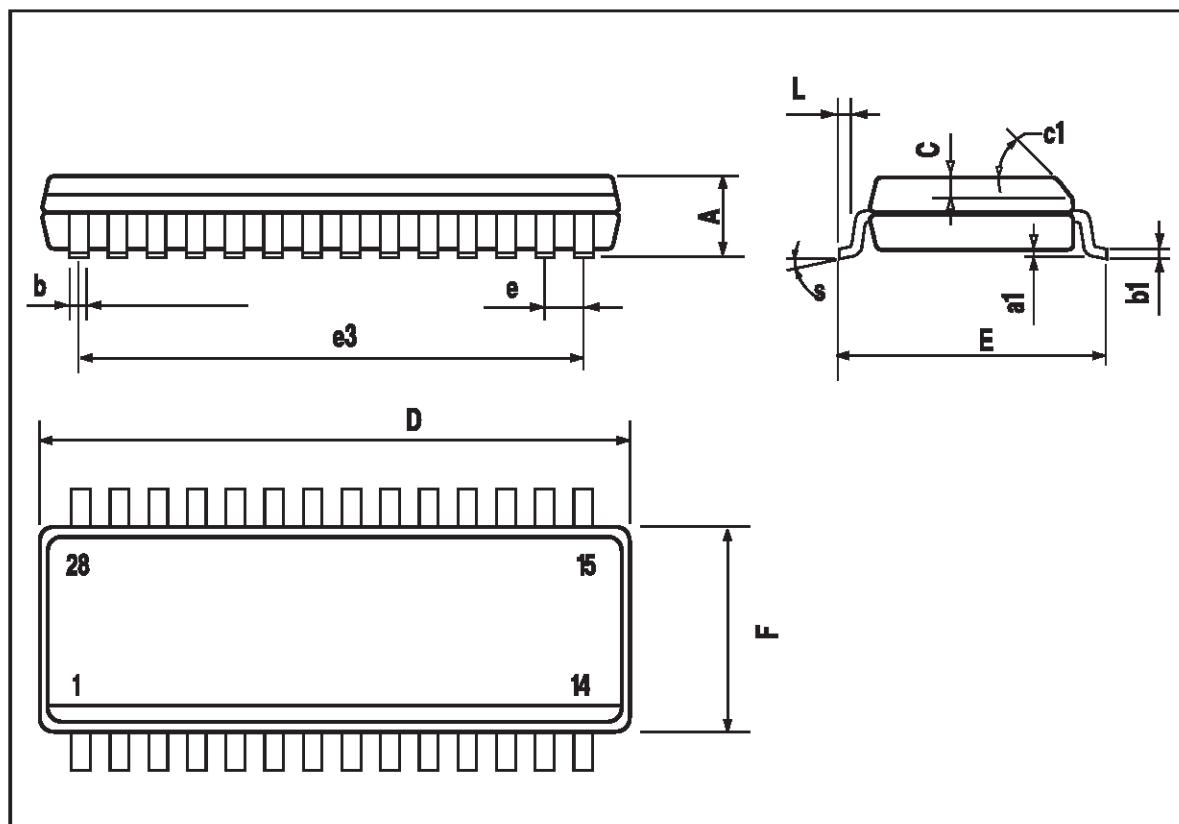
where  $\delta = t_p/T$

Thermal Parameter

Area/island ( $\text{cm}^2$ )	0.5	1	2	3	6
$R1=R5 (^{\circ}\text{C}/\text{W})$	0.22				
$R2=R6 (^{\circ}\text{C}/\text{W})$	3.5				
$R3=R7 (^{\circ}\text{C}/\text{W})$	26				
$R4=R8 (^{\circ}\text{C}/\text{W})$	62.28	52.28	44.28	40.28	32.28
$C1=C5 (\text{W.s}/^{\circ}\text{C})$	3.00E-03				
$C2=C6 (\text{W.s}/^{\circ}\text{C})$	2.50E-02				
$C3=C7 (\text{W.s}/^{\circ}\text{C})$	0.2				
$C4=C8 (\text{W.s}/^{\circ}\text{C})$	1.6	1.61	1.7	2	3.25
$R10=R11 (^{\circ}\text{C}/\text{W})$	150				

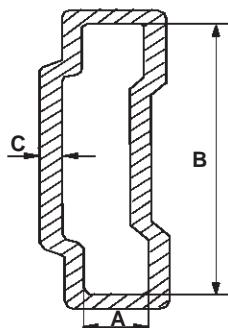
## SO-28 MECHANICAL DATA

DIM.	mm.			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			2.65			0.104
a1	0.10		0.30	0.004		0.012
b	0.35		0.49	0.013		0.019
b1	0.23		0.32	0.009		0.012
C		0.50			0.020	
c1		45 (typ.)				
D	17.7		18.1	0.697		0.713
E	10.00		10.65	0.393		0.419
e		1.27			0.050	
e3		16.51			0.650	
F	7.40		7.60	0.291		0.299
L	0.40		1.27	0.016		0.050
S		8 (max.)				



## VND920

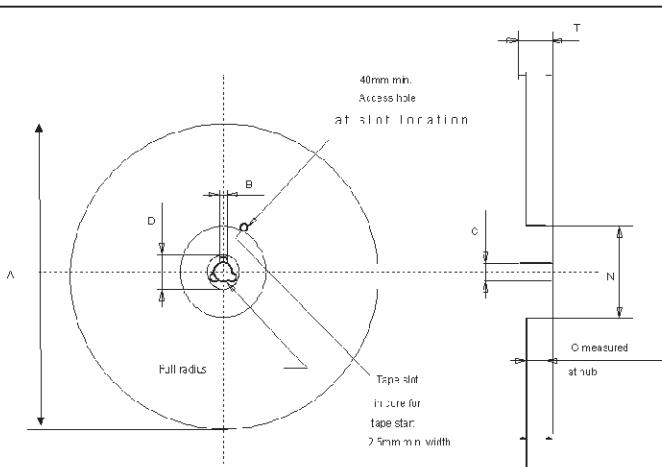
### SO-28 TUBE SHIPMENT (no suffix)



<b>Base Q.ty</b>	28
<b>Bulk Q.ty</b>	700
<b>Tube length (<math>\pm 0.5</math>)</b>	532
<b>A</b>	3.5
<b>B</b>	13.8
<b>C (<math>\pm 0.1</math>)</b>	0.6

All dimensions are in mm.

### TAPE AND REEL SHIPMENT (suffix "13TR")



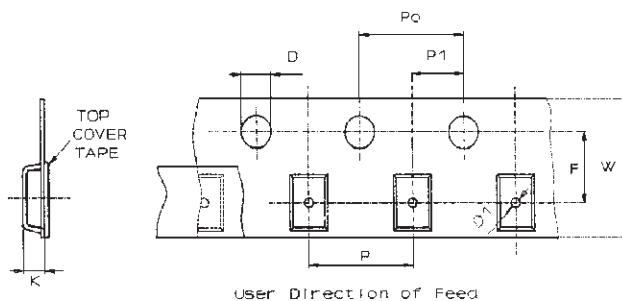
### REEL DIMENSIONS

<b>Base Q.ty</b>	1000
<b>Bulk Q.ty</b>	1000
<b>A (max)</b>	330
<b>B (min)</b>	1.5
<b>C (<math>\pm 0.2</math>)</b>	13
<b>F</b>	20.2
<b>G (+ 2 / -0)</b>	16.4
<b>N (min)</b>	60
<b>T (max)</b>	22.4

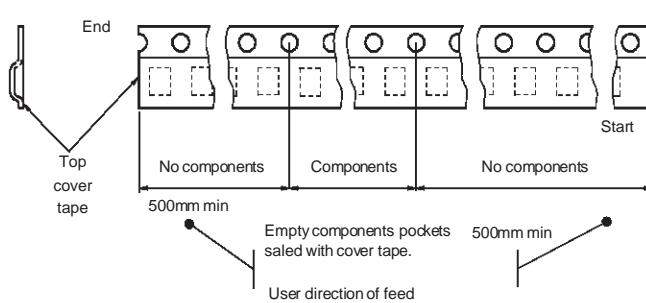
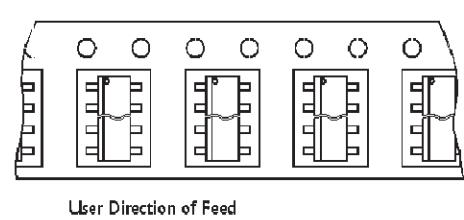
### TAPE DIMENSIONS

According to Electronic Industries Association (EIA) Standard 481 rev. A, Feb 1986

<b>Tape width</b>	<b>W</b>	16
<b>Tape Hole Spacing</b>	<b>P0 (<math>\pm 0.1</math>)</b>	4
<b>Component Spacing</b>	<b>P</b>	12
<b>Hole Diameter</b>	<b>D (<math>\pm 0.1/-0</math>)</b>	1.5
<b>Hole Diameter</b>	<b>D1 (min)</b>	1.5
<b>Hole Position</b>	<b>F (<math>\pm 0.05</math>)</b>	7.5
<b>Compartment Depth</b>	<b>K (max)</b>	6.5
<b>Hole Spacing</b>	<b>P1 (<math>\pm 0.1</math>)</b>	2



All dimensions are in mm.



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