



Features

- Two Identical Interface Channels
- Pre-regulated Smoothed Voltage and a Supply Current up to 50 mA for the Sensors
- Data from the Sensors by Current Modulation with a Transmission Rate of 60 kBaud (Transmission Bandwidth 500 kHz)
- TTL-compatible Input Activates the Sensor
- Data Output Can be Directly Connected to a Microcontroller Input
- Operation Supply Voltage Range $5.7\text{ V} \leq V_S \leq 40\text{ V}$
- ESD Protection According to MIL-STD-883C Test Method 3015.7
- High-level EMI Protection

Benefits

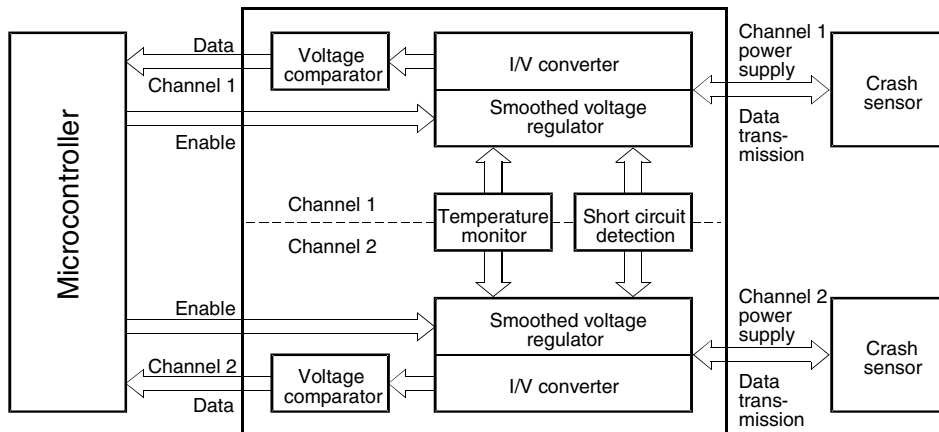
- Simple Wiring Thanks to One Common Line for Supply of the Sensor and Data Transmission from the Sensor to the U6268B
- Current Modulation Provides High Noise Immunity for Data Transfer

Description

The U6268B is an interface IC for remote automotive sensors. It links the crash sensors in the driver and passenger door with the main airbag unit in the dashboard. Two identical channels supply the external sensors and receive digital information from them via one active wire each. The interface supplies the external sensors with a pre-regulated smoothed voltage, the external units transmit the digital information back to the interface by current modulation.

As the device is designed for safety-critical applications, the highest data transmission security is mandatory. With high immunity against cross coupling between the two channels, the U6268B is tailored for the harsh automotive environment.

Figure 1. Block Diagram



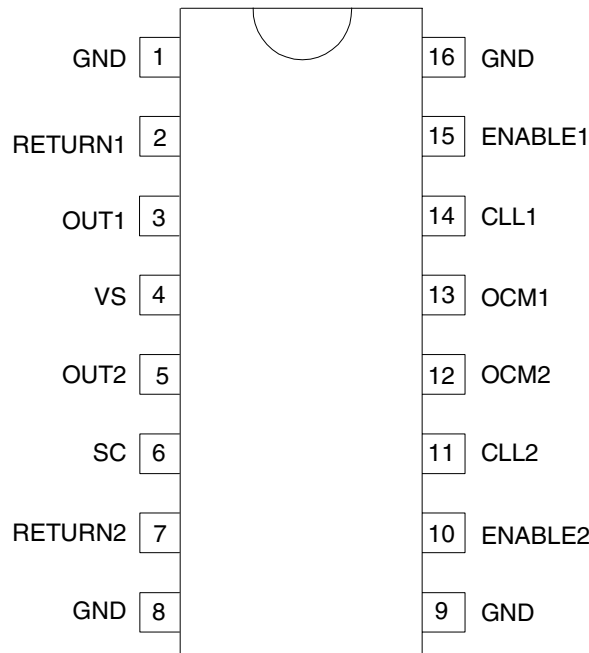
Side-airbag Sensor Dual Interface

U6268B



Pin Configuration

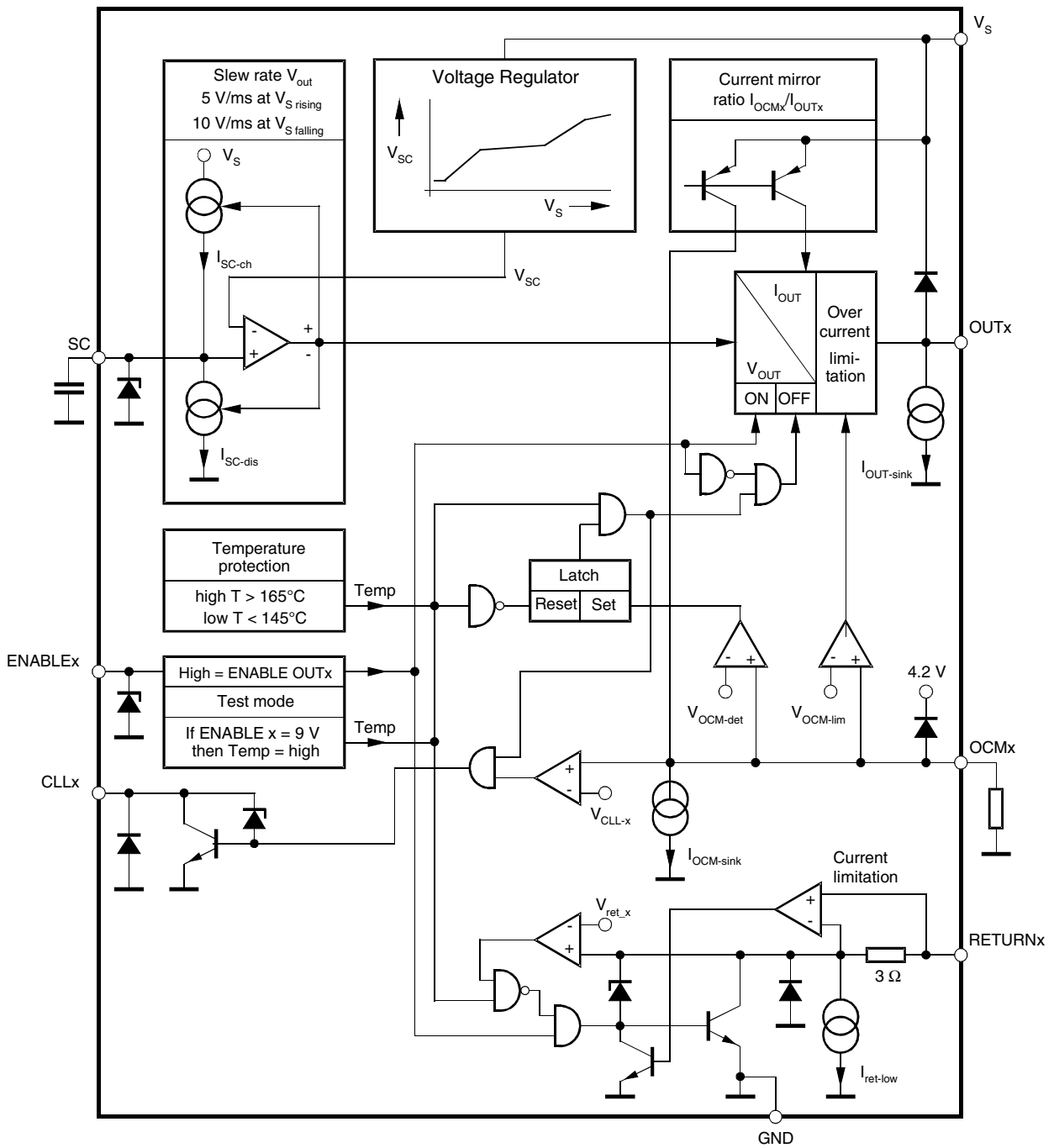
Figure 2. Pinning SO16



Pin Description

Pin	Symbol	Function
1	GND	Ground and reference pin
2	RETURN1	Return line of the external unit, internally connected to GND via a line-protection transistor
3	OUT1	Voltage-stabilized supply output and current-modulation input
4	VS	Supply voltage of the IC
5	OUT2	Voltage-stabilized supply output and current-modulation input
6	SC	Smooth time constant for slow voltage change at both OUT pins
7	RETURN2	Return line of the external unit, internally connected to GND via a line-protection transistor
8, 9	GND	Ground and reference pin
10	ENABLE2	Controls OUT1 voltage ENABLE1 High = OUT1 active, ENABLE1 Low or open = OUT1 switched off
11	CLL2	Current logic level output (low at high OUT2 current, monitoring via OCM2)
12	OCM2	Analog current output, representing 1/10 current of OUT2
13	OCM1	Analog current output, representing 1/10 current of OUT1
14	CLL1	Current logic level output (low at high OUT1 current, monitoring via OCM1)
15	ENABLE1	Controls OUT2 voltage ENABLE2 High = OUT2 active, ENABLE1 Low or open = OUT2 switched off
16	GND	Ground and reference pin

Figure 3. Functional Block Diagram



Functional Description

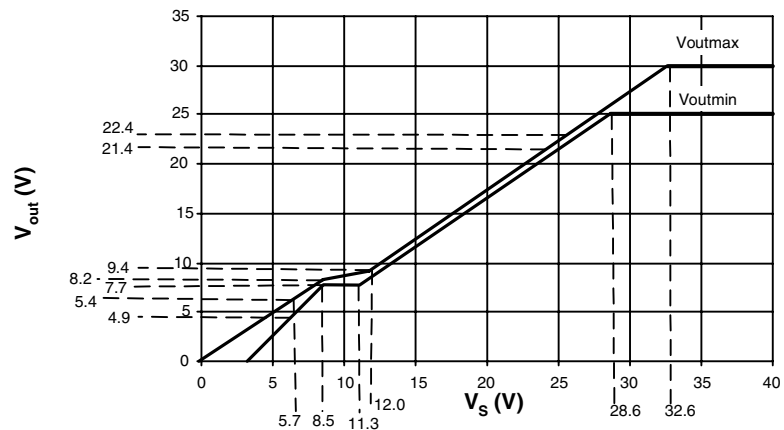
V_S

The IC and the external units are powered via the V_S pin 4. This pin is connected to the battery via a reverse battery protection diode. An electrolytic capacitor of 22 μF smoothes the voltage and absorbs positive and negative transients.

OUT1, OUT2

OUTx provides a smoothed, very slowly changing supply voltage for the external units and monitors the output current. During normal operating conditions, the OUTx voltage is typically 3 V below V_S , and changes very slowly with a varying battery voltage in order to suppress disturbances in the data transmission. At low V_S (5.7 to 8.5 V), the OUTx voltage is typically 0.5 V below V_S . This voltage difference is reduced in to ensure sufficient supply voltage for the external unit between OUTx and RETURNx. The output current capability is 50 mA. The internal pull-down current at OUTx is typically 3 mA.

Figure 4. Output Voltage with Tolerances versus Supply Voltage



The data transmission from the external unit to the interface IC is carried out on the same line by varying the current level. The quiescent current consumption of the external unit is about 5 to 15 mA. This current level is interpreted as logic high level at the CLL pin. The external unit can switch on an additional current of 30 mA, interpreted by the interface as logic low. The slope time of the current pulse is approximately 1 μs which is suitable for a transmission rate up to 60 kBaud. The necessary transmission bandwidth of greater than 500 kHz between OUTx and OCMx is guaranteed (see “Application Circuit” on page 12). To achieve good current transmission behavior, the dynamic resistance of OUTx may not exceed 12 Ω within the bandwidth range (total of 15 Ω for OUTx and RETURN).

The OUTx voltage can be switched off by ENABLEx = LOW to reset the external unit and to reduce power dissipation during fault conditions.

The OUT pins are protected against overtemperature and short circuits. A reverse polarity diode at pin V_S (pin 4) ensures that no current is fed back to the V_{Batt} system in the case of a short between OUTx and V_{Batt} . A minimum capacity of 33 nF is required at the pins OUTx.

ENABLE1, ENABLE2

ENABLE_x is a microcontroller-compatible input which switches the related output on or off.

- A low or open circuit applied to ENABLE_x switches off the related OUT_x and RETURN_x (high impedance). A sink current at pin OUT_x discharges the capacitive load.
- A high applied to ENABLE_x switches on the related OUT_x and RETURN_x to supply the external unit.

OCM1, OCM2

The output current of OUT_x is monitored with a transmission factor of 0.1 to the OCM_x. With a resistor from OCM to GND, the current is converted to a voltage. The electrical characteristics are specified by $R_{OCM} = 750 \Omega$. The CLL-current threshold, the OUT-current limitation and the OUT-current detection can be changed by varying R_{OCM} in a range of 500 Ω to 1 k Ω .

Current monitoring enables the device to detect overcurrent conditions at OUT_x (short-circuit to GND or RETURN_x) and low current conditions at OUT_x (short-circuit to V_{Batt} or open load).

The internal pull-down current at OUT_x creates no OCM_x-current. During ENABLE, the minimum voltage at OCM_x is the saturation voltage of an internal NPN-transistor with typically 0.1 V. The maximum voltage at OCM is limited by an internal clamping diode to 5.3 V.

CLL1, CLL2

The current at pin OUT_x is evaluated logically and ready to use for a microcontroller input. With this stage, the logic data transmission from the external unit to the interface is completed.

CLL_x is the output stage of a comparator with an internal threshold and with the OCM_x input. A OCM_x voltage higher than 2.4 V creates a logic low at CLL_x, and a OCM_x voltage lower than 1.43 V creates a logic high at CLL_x. The comparator has an internal hysteresis of typically 0.4 V.

With the pull-down resistor $R_{OCMx} = 750 \Omega$ at OCM_x, the correct OUT_x-current threshold related to the logical output CLL_x is ensured. The CLL_x is low if the OUT_x-current is higher than 27.3 mA, and the CLL_x is “high”, if the OUT_x-current is lower than 19.1 mA. The comparator has an internal hysteresis of typically 5 mA. The tolerance of the R_{OCM} resistor is assumed to be 0%.

The CLL pin is an open-collector output and needs a pull-up resistor of typically 2 k Ω to the 5-V supply. For ESD protection, a 7-V Zener diode is implemented.

RETURN 1, RETURN 2

The RETURN_x pin provides a low-ohmic connection to GND via a switched open-collector NPN-transistor. If ENABLE_x is high, RETURN_x is switched on with a saturation voltage of less than 0.5 V at $I_{RETURNx} \leq 50$ mA. If ENABLE_x is low or open, RETURN_x is a current sink with ≤ 2 mA. RETURN_x is current-limited at typically 150 mA.

SC

The smooth capacitor is designed to realize the long-time constant for the slow voltage change at OUT_x for both interface channels. The capacity is typically 22 nF. At the rising edge of V_{Batt} , the maximum slew rate is $V_{OUTx} = 5$ V/ms, and at the falling edge of V_{Batt} , the maximum slew rate is $V_{OUTx} = 10$ V/ms.

GND Pins

A GND bond from the chip to pin 1 and pin 8 provides high ground breakage security and the lowest voltage drop and ground shift between the IC and circuit ground. The four GND pins and the die pad are directly connected to the copper leadframe, resulting in a very low thermal resistance, R_{thJC} . To also achieve a low ambient thermal resistance (R_{thJA}) it is recommended metal parts of the housing be connected in a proper way with the GND pins.

Power Dissipation

Worst case calculation of the supply current I_S :

$$I_S = 1,278 \times (I_{OUT1} + I_{OUT2}) + 18 \text{ mA}$$

Worst case calculation of the IC's power dissipation P_V :

$$P_V = (V_S \times I_S) - [(V_S - V_{diff} - V_{ret-sat}) \times (I_{OUT1} + I_{OUT2}) + R_{OCM} \times ((I_{OUT1}^2 + I_{OUT2}^2)/81)]$$

V_S = Supply voltage (5.7 to 25 V)

V_{diff} = V_S to V_{OUTx} voltage difference

$$V_{diff} = 3.6 \text{ V at } 12 \text{ V} \leq V_S \leq 25 \text{ V}$$

$$V_{diff} = 0.8 \text{ V at } 5.7 \text{ V} \leq V_S \leq 8.5 \text{ V}$$

$V_{ret-sat}$ = 0.5 V saturation voltage return

I_{OUTx} = output current at pin $OUTx$ = 0 to 60 mA

R_{OCM} = resistor at pin $OCMx$

An integrated overtemperature protection generates a switch-off signal at a chip temperature of typically $T_j = 160^\circ\text{C}$ and a switch-on signal at typically $T_j = 150^\circ\text{C}$.

If overtemperature is detected, only the corresponding channel will be disabled. The other channel stays enabled.

The RETURNx is switched off if the voltage at RETURNx exceeds 2 V (short-circuit comparator threshold) and overtemperature is detected.

The OUTx is switched off if the voltage at OCMx is higher than 4.6 V (overcurrent detection level) and overtemperature is detected. The OCM voltage monitors the output current at OUTx via the current ratio of 0.1. The overcurrent-detection level of OUTx can be varied by changing the OCMx resistor. If OUTx is switched off by overtemperature and overcurrent detection, the CLLx output remains a logic low (overcurrent).

As the IC is only overtemperature-protected against short-circuit conditions at RETURNx or OUTx, it has to be checked in each application that the chip temperature does not exceed $T_{jmax} = 150^\circ\text{C}$ in normal operation.

Test Hint

The overtemperature signal can be activated by connecting ENABLE1 or ENABLE2 to 9 V/10 mA.

Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameters	Symbol	Min.	Max.	Unit
Supply voltage	V_S	-0.6	40	V
Voltage at pins CLL1, CLL2, ENABLE1, ENABLE2		-0.3	6	V
Voltage at SC	V_{SC}	-0.3	30	V
Voltage at OCM1, OCM2	V_{OCMx}	-0.3	6.8	V
Voltage at RETURN1, RETURN2	$V_{RETURNx}$	-1	27	V
Voltage at OUT1, OUT2	V_{OUTx}	-1	40	V
Current at supply (both channels OUTx and RETURNx shorted)	I_S		240	mA
Current at logical pins: CLL1, CLL2 ENABLE1, ENABLE2	I_{CCLx} $I_{ENABLEx}$		3 0.1	mA mA
Current at SC (SC related to GND or V_{Batt})	I_{SC}	-110	220	μ A
Current at pins to external unit OUT1, OUT2, RETURN1, RETURN2		Internally limited		
ESD classification Human body model (100 pF, 1.5 k Ω) Machine model (200 pF, 0.0 Ω)	All pins	\pm 2000 \pm 200		V V
Ambient temperature range	T_{amb}	-40	95	$^{\circ}$ C
Junction temperature range	T_j	-40	150	$^{\circ}$ C
Storage temperature range	T_{stg}	-55	125	$^{\circ}$ C

Thermal Resistance⁽¹⁾

Parameters	Symbol	Value	Unit
Junction case	R_{thJC}	36	K/W

Note: 1. A good ambient thermal resistance junction ($R_{thJA} = 65$ K/W) can be achieved by using a big pad size for ground connection near a metal component (see section “GND Pins” on page 6)

Electrical Characteristics

$T_{amb} = -40^{\circ}\text{C}$ to 95°C and $T_j = -40^{\circ}\text{C}$ to 150°C ,

Operation supply-voltage range $V_S = 5.7$ to 18 V continuously, $V_S \leq 25$ V for maximum 25 min, $V_S \leq 40$ V for up to 500 ms.

The current values are based on $R = 750 \Omega$, 0%-resistor at OCM1/OCM2 pins.

Parameters	Test Conditions	Symbol	Min.	Typ.	Max.	Unit
Supply Current $T_j \geq 125^{\circ}\text{C}$	Outputs disabled, $V_S \leq 18$ V	I_S			8	mA
	Outputs disabled, $V_S \leq 40$ V	I_S			14	mA
	One output enabled, $V_S \leq 18$ V	I_S			13	mA
	Both outputs enabled, $V_S \leq 18$ V	I_S			18	mA
	Output load 2×15 mA, $V_S \leq 18$ V	I_S			56	mA
	Output load 2×28 mA, $V_S \leq 18$ V	I_S			90	mA
	Output load 2×50 mA, $V_S \leq 18$ V	I_S			146	mA
	Output load 2×60 mA, $V_S \leq 18$ V ($T_j > 125^{\circ}\text{C}$)	I_S			171	mA
	Both channels OUTx and RETURNx shorted, $V_S \leq 18$ V	I_S			200	mA
Function SC						
Voltage at SC	$V_S = 5.7$ V	V_{SC}	5.1		5.3	V
Voltage at SC	$V_S = 12.5$ V	V_{SC}	9		9.4	V
Maximal voltage at SC	$V_S = 40$ V	V_{SCmax}			30	V
SC-discharge current	Voltage SC = $V_{SC} - 3$ V 5.7 V $\leq V_S \leq 40$ V	I_{SC_dis}	33		82	μA
SC-charge current	Voltage SC = $V_{SC} - 3$ V 5.7 V $\leq V_S \leq 40$ V	I_{SC_ch}	-58		-20	μA
Function OUT1 and OUT2 (See Figure 4 on page 4)						
Voltage difference, V_S to V_{OUTx}	$I_{OUTx} = 5$ to 50 mA 5.7 V $\leq V_S \leq 8.5$ V 12 V $\leq V_S \leq 25$ V	V_{diff_low}	0.3		0.8	V
		V_{diff_high}	2.6		3.6	V
Output voltage OUTx	8.5 V $\leq V_S \leq 11.3$ V	V_{OUT_med}	7.7			V
Maximal voltage at OUTx	$V_S = 40$ V	V_{OUT_max}	25		30	V
Current mirror ratio, I_{OCMx}/I_{OUTx}	$V_S \leq 40$ V, $I_{OUTx} = 5$ to 15 mA $V_S \leq 25$ V, $I_{OUTx} = 15$ to 50 mA $V_S \leq 40$ V, $I_{OUTx} = 15$ to 50 mA	I_{OUT_ratio}	0.09		0.12	
			0.10		0.11	
			0.097		0.11	
Linearity of mirror ratio I_{OCMx}/I_{OUTx}		Ratio_lin	-5		5	%
Dynamic resistance OUTx	$V_S \leq 40$ V $I_{OUT} = 15$ to 50 mA	R_{OUT}	2		12	Ω
Dynamic resistance OUTx + RETURNx	$V_S \leq 40$ V $I_{OUT} = 15$ to 50 mA	R_{Dyn}	4		15	Ω
OUTx current limitation (OUTx short to GND)	$V_S \leq 18$ V $V_S \leq 40$ V	I_{OUT_lim}	-80		-60	mA
			-105		-60	mA
Overcurrent detection level	$T_j < 125^{\circ}\text{C}$	I_{OUT_det}	-70		-51	mA
	$T_j \geq 125^{\circ}\text{C}$ Always valid: current limitation is higher than overcurrent detection	I_{OUT_det}	-60		-51	mA
Maximum OUTx current (OUTx short to GND)	$V_S = 14$ V, OCMx shorted to GND	I_{OUT_max}	-140		-85	mA

Electrical Characteristics (Continued)

$T_{amb} = -40^{\circ}\text{C}$ to 95°C and $T_j = -40^{\circ}\text{C}$ to 150°C ,

Operation supply-voltage range $V_S = 5.7$ to 18 V continuously, $V_S \leq 25$ V for maximum 25 min, $V_S \leq 40$ V for up to 500 ms.

The current values are based on $R = 750 \Omega$, 0%-resistor at OCM1/OCM2 pins.

Parameters	Test Conditions	Symbol	Min.	Typ.	Max.	Unit
Leakage current at disabled OUTx	OUTx short to GND, $V_S \leq 25$ V OUTx short to GND, $V_S \leq 38.5$ V	I_{OUT_leak}	-0.02 -12			mA mA
Leakage voltage at disabled OUTx	OUTx open, $V_S \leq 38.5$ V	V_{OUT_leak}			4.3	V
Internal pull-down current	$V_S \leq 18$ V $V_S \leq 40$ V	I_{OUT_sink}	1.8 2.5		4 4.5	mA mA
Supply rejection ratio	$V_{SC} = 7.6$ V	V_{rej_mV}			80	mV
Supply rejection ratio	Variation of $V_S = 8.4$ V to 40 V in 10 ms	V_{rej_dB}	51.9			dB
Minimum capacity at OUTx for phase margin		C_{OUT_min}	33			nF
Delay time with $C_{out} = 47$ nF	Switching on ENABLE = 1 to 90% V_{OUT} reached	Enable_on	3		30	μs
	Switching off ENABLE = 0 to 10% V_{OUT} reached	Enable_off	30		100	μs
Function OCM1, OCM2						
Voltage threshold CLL-comparator	CLLx low-level voltage threshold	V_{CLL_L}	1.75		2.4	V
	CLLx high-level voltage threshold	V_{CLL_H}	1.43		1.9	V
	Voltage hysteresis	V_{CLL_hys}	0.26		0.6	V
Minimum voltage at OCMx	$I_{OUT} = 0$ to 5 mA	V_{OCM_min}			0.5	V
Current-limitation level	$V_S \leq 40$ V, OUTx short to GND	V_{OCM_lim}	4.3		5.3	V
Overcurrent-detection level	$V_S \leq 40$ V	V_{OCM_det}	4.2		4.9	V
Current limitation minus overcurrent detection	$V_{OCM_lim} - V_{OCM_over}$	Δ_lim_OCM	0.15		0.5	V
Internal pull-down current		I_{OCM_sink}	0.1		0.45	mA
Function RETURN1, RETURN2						
Enable high saturation voltage	$I_{RETURN} = 50$ mA	V_{ret_sat}			0.5	V
Dynamic resistance	$dI \geq 10$ mA	R_{ret}	2		8	Ω
Current limitation RETURNx is always higher than current limitation OUTx	Enable high, $V_{RETURNx} = 2$ V	I_{ret_lim}	60		150	mA
	Enable high, $V_{RETURNx} \leq 18$ V	I_{ret_lim}	70		200	mA
	Enable low, $V_{RETURNx} \leq 18$ V	I_{ret_low}	0.8		2	mA
Overcurrent-detection level	Threshold comparator, switch-off return	V_{ret_low}	1.4		2	V
	Threshold comparator, switch-on return	V_{ret_high}	1.1		1.5	V
	Hysteresis	V_{ret_hys}	0.2		0.7	V
Delay time $C_{RETURN} = 47$ nF	Switching on I_{RETURN} at 50 mA	t_{dRet_on}	3		30	μs
	Switching off I_{RETURN} at 1 mA	t_{dRet_off}	30		90	μs
Function CLL1, CLL2 (CLLx with 2 kΩ to 5 V)						
I_{OUT} threshold CLL comparator	$R_{OCM} = 750 \Omega$ CLL low-level threshold	I_{CLL_L}	23.3		27.3	mA
	CLL high-level threshold	I_{CLL_H}	19.1		22.3	mA
	Hysteresis	I_{CLL_hys}	3.5		8.2	mA
CLL saturation voltage	$I_{CLL} \leq 2.5$ mA	V_{CLL_sat}			0.4	V

Electrical Characteristics (Continued)

$T_{amb} = -40^{\circ}\text{C}$ to 95°C and $T_j = -40^{\circ}\text{C}$ to 150°C ,

Operation supply-voltage range $V_S = 5.7$ to 18 V continuously, $V_S \leq 25$ V for maximum 25 min, $V_S \leq 40$ V for up to 500 ms.

The current values are based on $R = 750 \Omega$, 0%-resistor at OCM1/OCM2 pins.

Parameters	Test Conditions	Symbol	Min.	Typ.	Max.	Unit
CLL leakage current	$V_{CLL} \leq 6.5$ V	I_{CLL_leak}			1	μA
Response time to current change	I_{OUT} to CLL rise	t_{CLL_rise}	0.1		2	μs
	I_{OUT} to CLL fall	t_{CLL_fall}	0.1		2	μs
	Maximum difference between rise and fall time	$t_{\Delta\text{-rise-fall}}$			1	μs
CLL output switching speed	Rise	t_{CLL_rise}			1	μs
	Fall	t_{CLL_fall}			1	μs
Current transmission rate			60			kHz
Current transmission 3 dB bandwidth			500			kHz
Function ENABLE1, ENABLE2						
Enable high-level threshold		V_{Enable_on}	2		6.5	V
Enable low-level threshold		V_{Enable_off}	-0.3		0.8	V
Enable input pull-down current (to ensure output disabled during power-off and reset of microcontroller)		I_{Enable}	10		100	μA
Power Dissipation						
Power dissipation 1 $T_j \geq 125^{\circ}\text{C}$	$V_S = 18$ V, $I_{OUT1} = 28$ mA, I_{OUT2} at overcurrent detection level or $I_{OUT2} = 28$ mA, I_{OUT1} at overcurrent detection level	P_{dis1}			1	W
Power dissipation 2 $T_j \geq 125^{\circ}\text{C}$	$V_S = 18$ V, $I_{OUT1} = I_{OUT2} = 28$ mA	P_{dis2}			0.75	W
Selective Overtemperature Protection						
Logic AND connected with overcurrent detection (RETURNx, OUTx)	Switch off	Temp_off	155		165	$^{\circ}\text{C}$
	Switch on	Temp_on	145		155	$^{\circ}\text{C}$
	Hysteresis	Temp_hys	5		20	$^{\circ}\text{C}$
Time delay until overtemperature shut-down	$V_S = 25$ V, $T_{amb} = 125^{\circ}\text{C}$ OUT1 = OUT2 = GND	t_{del}	100			ms

Timing Diagrams

Figure 5. Variation of Power Supply

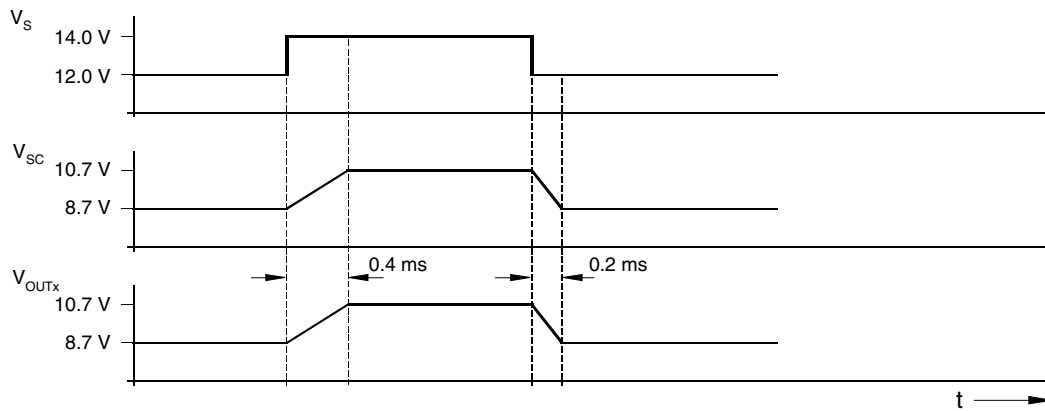


Figure 6. Overcurrent Protection

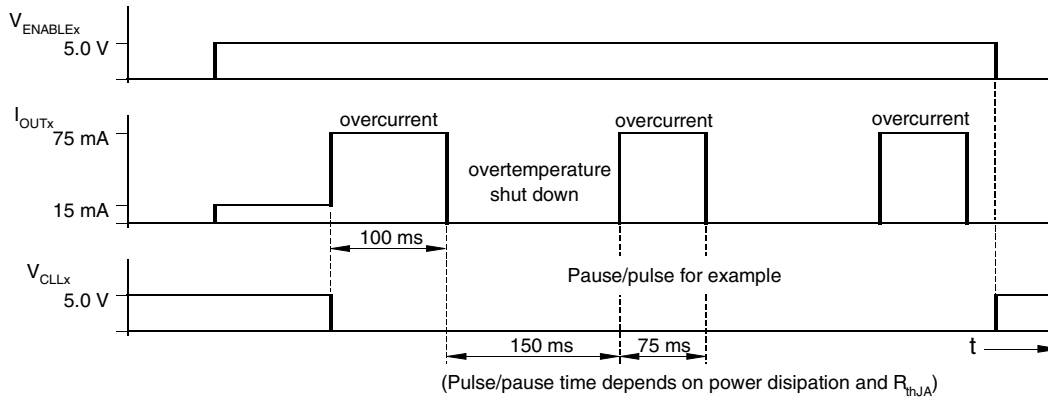


Figure 7. Data Transmission

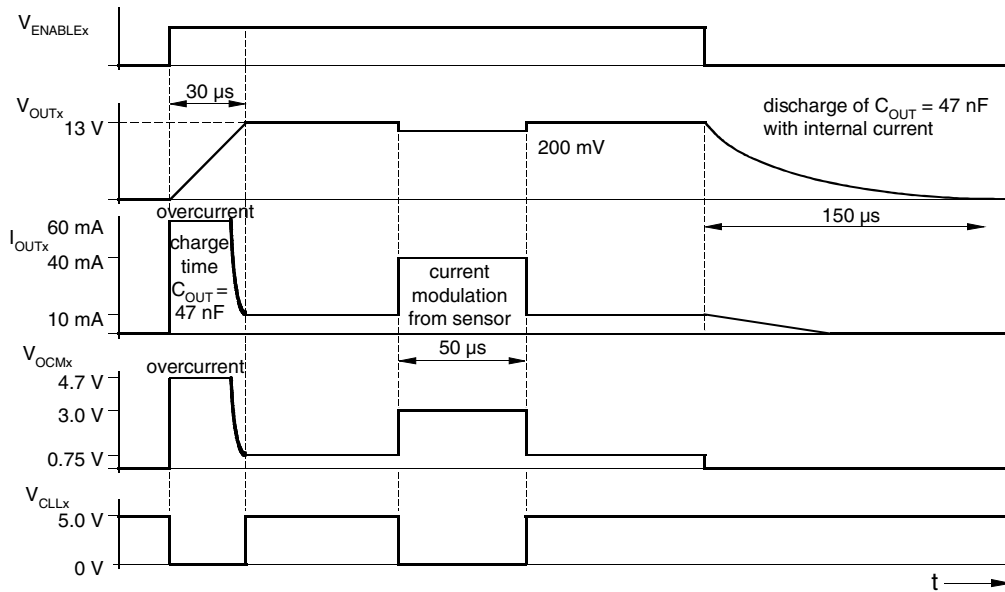
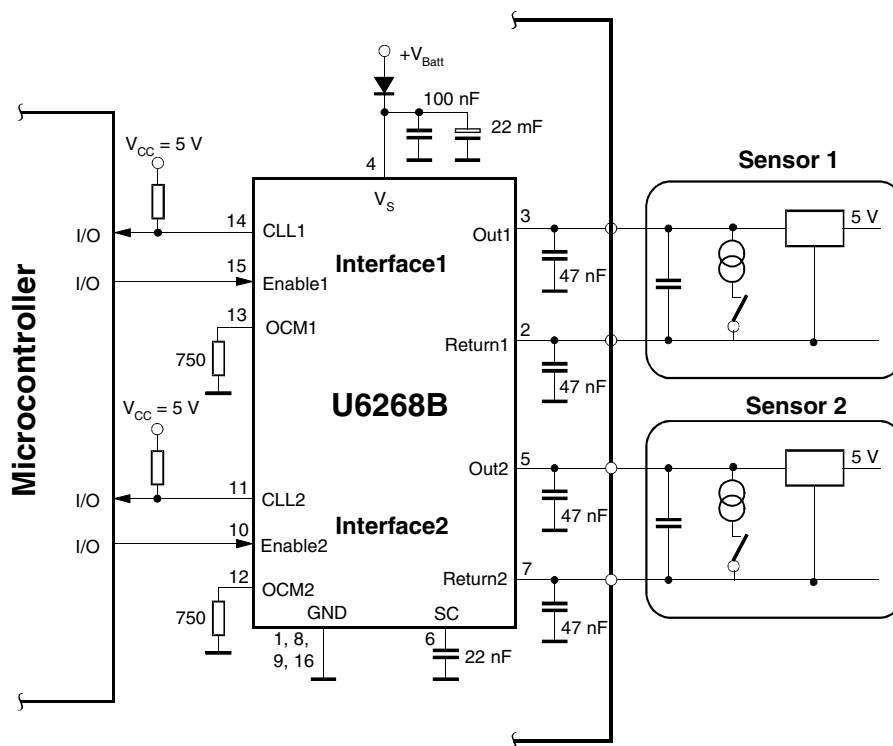


Figure 8. Application Circuit

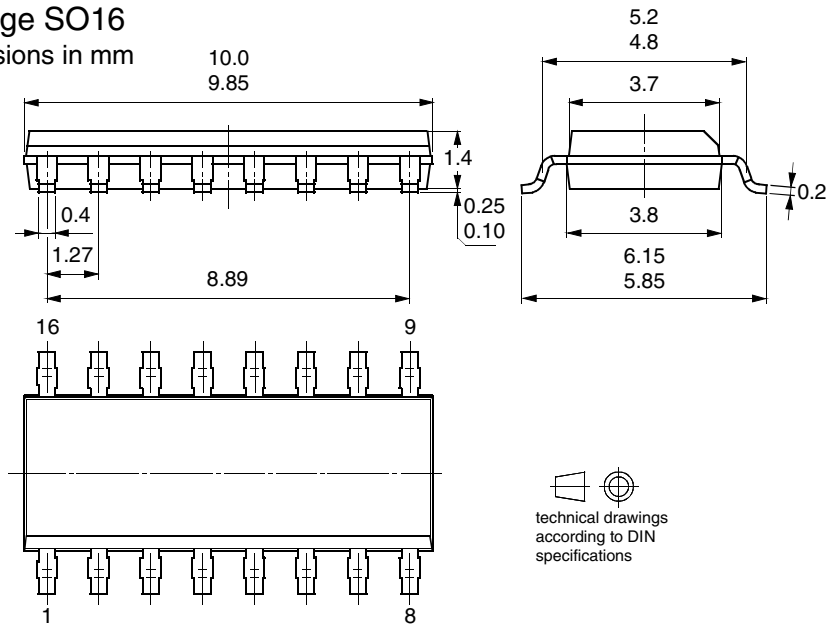


Ordering Information

Extended Type Number	Package	Remarks
U6268B-FP	SO16	–

Package Information

Package SO16
Dimensions in mm





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4808A-AUTO-06/04