Micropower 500-mA CMOS LDO Regulator With Error Flag/Power-On-Reset

FEATURES

- Input Voltage 2 V to 6 V
- Low 150-mV Dropout at 500-mA Load
- Guaranteed 500-mA Output Current
- Uses Low ESR Ceramic Output Capacitor
- Fast Load and Line Transient Response
- Only 100-μV(rms) Noise With Noise Bypass Capacitor
- 1-μA Maximum Shutdown Current
- Built-in Short Circuit and Thermal Protection
- Out-Of-Regulation Error Flag (Power Good or POR)
- Fixed 1.215-V, 1.5-V, 1.8-V, 2.0-V, 2.5-V, 2.8-V, 2.9-V, 3.0-V, 3.3-V, 5.0-V, or Adjustable Output Voltage Options

Pb-free Available

- Other Output Voltages Available by Special Order
- 1.1-W Power Dissipation
- Thin, Thermally Enhanced MLP33 PowerPAK® Package

APPLICATIONS

- Laptop and Palm Computers
- Desktop Computers
- Cellular Phones
- PDA, Digital Still Cameras

DESCRIPTION

The Si9185 is a 500-mA CMOS LDO (low dropout) voltage regulator. The device features ultra low ground current and dropout voltage to prolong battery life in portable electronics. The Si9185 offers line/load transient response and ripple rejection superior to that of bipolar or BiCMOS LDO regulators, and is designed to drive lower cost ceramic, as well as tantalum, output capacitors. An external noise bypass capacitor connected to the device's C_{NOISE} pin will lower the LDO's output noise for low noise applications. The Si9185 also includes an out-of-regulation error flag. If a capacitor is connected to the device's delay pin, the error flag output pin will generate a delayed power-on-reset signal. The device is guaranteed stable from maximum load current down to 0-mA load.

The Si9185 is available in both standard and lead (Pb)-free MLP33 PowerPAK packages and is specified to operate over the industrial temperature range of –40 °C to 85 °C. MLP33 PowerPAK packaging allows enhanced heat transfer to the PC board.

TYPICAL APPLICATIONS CIRCUITS

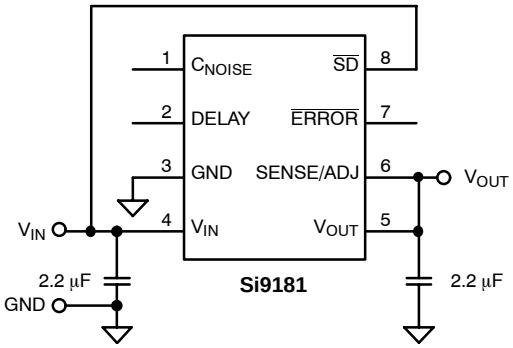


FIGURE 1. Fixed Output

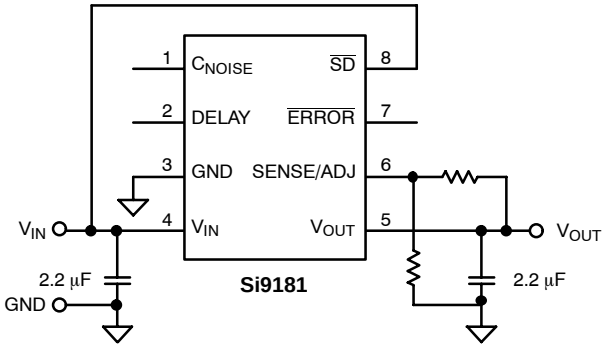


FIGURE 2. Adjustable Output

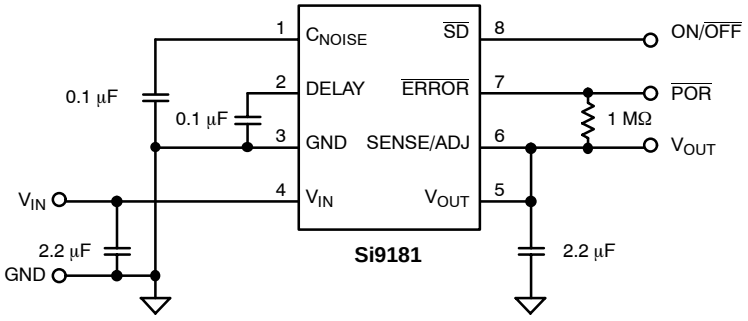


FIGURE 3. Low Noise, Full Features Application

ABSOLUTE MAXIMUM RATINGS

Input Voltage, V_{IN} . . . . . 6.5 V
$\overline{SD}$ Input Voltage, $V_{\overline{SD}}$ . . . . . −0.3 V to V_{IN}
Output Current, I_{OUT} . . . . . 500 mA Continuous, Short Circuit Protected
Output Voltage, V_{OUT} . . . . . −0.3 V to $V_{O(nom)}$ + 0.3 V
Maximum Junction Temperature, $T_{J(max)}$ . . . . . 150°C
Storage Temperature, T_{STG} . . . . . −55°C to 150°C
ESD (Human Body Model) . . . . . 2 kV

Power Dissipation[b] . . . . . 2.5 W
Thermal Impedance (Θ_{JA})[a]
($R\Theta_{JA}$) . . . . . 50°C/W
($R\Theta_{Jc}$) . . . . . 4°C/W

Notes
a. Device mounted with all leads soldered or welded to PC board. (PC board—2" x 2", 4-layer, FR4, 0.25 square inch spreading copper)
b. Derate 20 mW/°C above T_A = 25°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING RANGE

Input Voltage, V_{IN} . . . . . 2 V to 6 V
Output Voltage, V_{OUT} (Adjustable Version) . . . . . 1.215 V to 5 V
R2 . . . . . 25 kΩ to 150 kΩ

Operating Ambient Temperature, T_A . . . . . −40°C to 85°C
Operating Junction Temperature, T_J . . . . . −40°C to 125°C

C_{IN} = 2.2 μF, C_{OUT} = 2.2 μF (ceramic, X5R or X7R type) , C_{NOISE} = 0.1 μF (ceramic)
C_{OUT}Range = 1 μF to 10 μF (±10%, x5R or x7R type)
$C_{IN} \geq C_{OUT}$

SPECIFICATIONS

Parameter	Symbol	Test Conditions Unless Otherwise Specified $V_{IN} = V_{OUT(nom)} + 1$ V, I_{OUT} = 1 mA C_{IN} = 2.2 μF, C_{OUT} = 2.2 μF, $V_{\overline{SD}}$ = 1.5 V	Temp[a]	Limits −40 to 85°C Min[b]	Typ[c]	Max[b]	Unit
Output Voltage Range	V_{OUT}	Adjustable Version	Full	1.215		5	V
Output Voltage Accuracy (Fixed Versions)		1 mA ≤ I_{OUT} ≤ 500 mA	Room	−1.5		1.5	% $V_{O(nom)}$
			Full	−2.5		2.5	
Feedback Voltage (ADJ Version)	V_{ADJ}		Room	1.191	1.215	1.239	V
			Full	1.179		1.251	
Line Regulation ($V_{ADJ} \leq V_{OUT} \leq$ 4 V)	$\frac{\Delta V_{OUT} \times 100}{V_{IN} \times V_{OUT}}$	From $V_{IN} = V_{OUT}$ + 1 V to V_{OUT} + 2 V	Full	−0.18		0.18	%/V
Line Regulation (4 V $V_{OUT} \leq$ 5 V)		From V_{IN} = 5.5 V to 6 V	Full	−0.18		0.18	

SPECIFICATIONS

Parameter	Symbol	Test Conditions Unless Otherwise Specified $V_{IN} = V_{OUT(nom)} + 1$ V, $I_{OUT} = 1$ mA $C_{IN} = 2.2$ µF, $C_{OUT} = 2.2$ µF, $V_{\overline{SD}} = 1.5$ V		Temp[a]	Limits –40 to 85°C Min[b]	Typ[c]	Max[b]	Unit
Dropout Voltage[d] (@$V_{OUT(nom)} \geq 2$ V)	$V_{IN} - V_{OUT}$	I_{OUT} = 10 mA		Room		5	20	mV
		I_{OUT} = 200 mA		Room		145	215	
		I_{OUT} = 500 mA		Room		320	480	
				Full			600	
Dropout Voltage[d] (@$V_{OUT(nom)} \geq 2.5$ V)		I_{OUT} = 200 mA		Room		115	175	
		I_{OUT} = 500 mA		Room		250	400	
				Full			480	
Dropout Voltage[d] (@$V_{OUT(nom)} \geq 3.3$ V)	$V_{IN} - V_{OUT}$	I_{OUT} = 200 mA		Room		90	135	
		I_{OUT} = 500 mA		Room		200	300	
				Full			400	
Dropout Voltage[d] (@$V_{OUT(nom)} \geq 5$ V)	$V_{IN} - V_{OUT}$	I_{OUT} = 200 mA		Room		60	100	
		I_{OUT} = 500 mA		Room		150	210	
				Full			300	
Dropout Voltage[d] (@$V_{OUT(nom)} < 2$ V, $V_{IN} \geq 2$ V)	$V_{IN} - V_{OUT}$	I_{OUT} = 200 mA		Room		170	250	
		I_{OUT} = 500 mA		Room		415	625	
				Full			825	
Ground Pin Current	I_{GND}	I_{OUT} = 0 mA		Room		150		µA
		I_{OUT} = 200 mA		Room		1000		
				Full			1500	
		I_{OUT} = 500 mA		Room		2500		
				Full			4000	
Shutdown Supply Current	$I_{IN(off)}$	$V_{\overline{SD}} = 0$ V		Room		0.1	1	µA
ADJ Pin Current	I_{ADJ}	ADJ = 1.2 V		Room		5	100	nA
Peak Output Current	$I_{O(peak)}$	$V_{OUT} \geq 0.95 \times V_{OUT(nom)}$, $t_{pw} = 2$ ms		Room	600			mA
Output Noise Voltage	e_N	BW = 50 Hz to 100 kHz I_{OUT} = 150 mA	w/o C_{NOISE}	Room		200		µV (rms)
			C_{NOISE} = 0.1 µF	Room		100		
Ripple Rejection	$\Delta V_{OUT}/\Delta V_{IN}$	I_{OUT} = 150 mA	f = 1 kHz	Room		60		dB
			f = 10 kHz	Room		60		
			f = 100 kHz	Room		40		
Dynamic Line Regulation	$\Delta V_{O(line)}$	V_{IN} : $V_{OUT(nom)}$ + 1 V to $V_{OUT(nom)}$ + 2 V t_R/t_F = 5 µs, I_{OUT} = 500 mA		Room		10		mV
Dynamic Load Regulation	$\Delta V_{O(load)}$	I_{OUT} : 1 mA to 150 mA, t_R/t_F = 2 µs		Room		30		
V_{OUT} Turn-On-Time	t_{ON}	V_{IN} = 4.3 V V_{OUT} = 3.3 V	w/o C_{NOISE} Cap	Room		5		µs
			C_{NOISE} = 0.1 µF	Room		2		mS
Thermal Shutdown								
Thermal Shutdown Junction Temp	$t_{J(s/d)}$			Room		165		°C
Thermal Hysteresis	t_{HYST}			Room		20		
Short Circuit Current	I_{SC}	V_{OUT} = 0 V		Room		800		mA
Shutdown Input								
$\overline{SD}$ Input Voltage	V_{IH}	High = Regulator ON (Rising)		Full	1.5		V_{IN}	V
	V_{IL}	Low = Regulator OFF (Falling)		Full			0.4	

SPECIFICATIONS

Parameter	Symbol	Test Conditions Unless Otherwise Specified $V_{IN} = V_{OUT(nom)} + 1$ V, $I_{OUT} = 1$ mA $C_{IN} = 2.2$ µF, $C_{OUT} = 2.2$ µF, $\overline{SD} = 1.5$ V	Temp[a]	Limits −40 to 85°C Min[b]	Typ[c]	Max[b]	Unit
$\overline{SD}$ Input Current[e]	I_{IH}	$V_{\overline{SD}} = 0$ V, Regulator OFF	Room		0.01		µA
	I_{IL}	$V_{\overline{SD}} = 6$ V, Regulator ON	Room		1.0		
Shutdown Hysteresis	V_{HYST}		Full		100		mV
Error Output							
Output High Leakage	I_{OFF}	$\overline{ERROR} = V_{OUT(nom)}$	Full		0.01	2	µA
Output Low Voltage[g]	V_{OL}	$I_{SINK} = 2$ mA	Full			0.4	V
Out-of-Regulation Error Flag Threshold Voltage (rising)[g]	V_{TH}		Full	0.93 x V_{OUT}	0.95 x V_{OUT}	0.97 x V_{OUT}	
Hysteresis[g]	V_{HYST}		Room		2% x V_{OUT}		
Delay Pin Current Source	I_{DELAY}		Room	1.2	2.2	3.0	µA

Notes
a. Room = 25°C, Full = −40 to 85°C.
b. The algebraic convention whereby the most negative value is a minimum and the most positive a maximum.
c. Typical values are for DESIGN AID ONLY, not guaranteed nor subject to production testing. Typical values for dropout voltage at $V_{OUT} \geq 2$ V are measured at $V_{OUT} = 3.3$ V, while typical values for dropout voltage at $V_{OUT} < 2$ V are measured at $V_{OUT} = 1.8$ V.
d. Dropout voltage is defined as the input to output differential voltage at which the output voltage drops 2% below the output voltage measured with a 1-V differential, provided that V_{IN} does not not drop below 2.0 V. When $V_{OUT(nom)}$ is less than 2.0 V, the output will be in regulation when 2.0 V − $V_{OUT(nom)}$ is greater than the dropout voltage specified.
e. The device's shutdown pin includes a typical 6-MΩ internal pull-down resistor connected to ground.
f. V_{OUT} is defined as the output voltage of the DUT at 1 mA.
g. The Error Output (Low) function is guaranteed for $V_{IN} \geq 2.0$ V.

TIMING WAVEFORMS

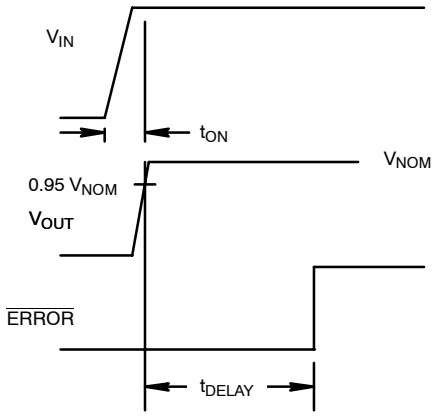
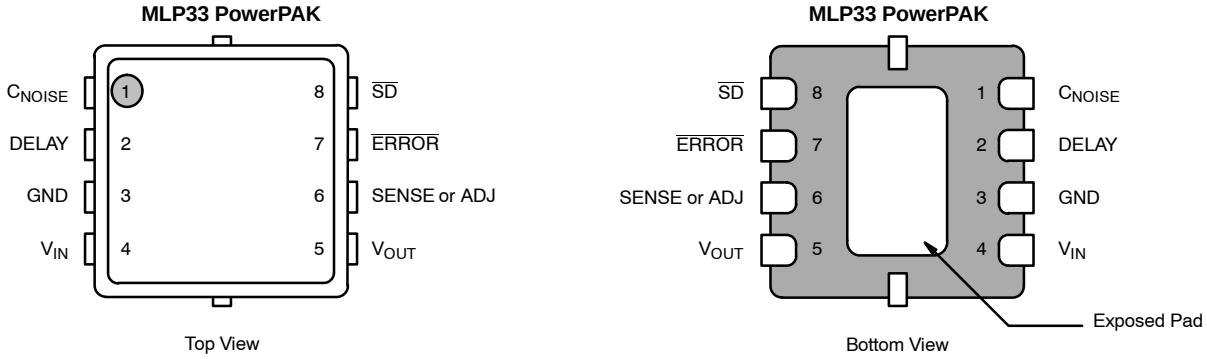


FIGURE 4. Timing Diagram for Power-Up

PIN CONFIGURATION



PIN DESCRIPTION

Pin Number	Name	Function
1	C_{NOISE}	Noise bypass pin. For low noise applications, a 0.01-µF or larger ceramic capacitor should be connected from this pin to ground.
2	DELAY	Capacitor connected from this pin to ground will allow a delayed power-on-reset signal at the $\overline{ERROR}$ (Pin 7) output. Refer to Figure 4.
3	GND	Ground pin. Local ground for C_{NOISE} and C_{OUT}.
4	V_{IN}	Input supply pin. Bypass this pin with a 2.2-µF ceramic or tantalum capacitor to ground.
5	V_{OUT}	Output voltage. Connect C_{OUT} between this pin and ground.
6	SENSE or ADJ	For fixed output voltage versions, this pin should be connected to V_{OUT} (Pin 5). For adjustable output voltage version, this voltage feedback pin sets the output voltage via an external resistor divider.
7	$\overline{ERROR}$	This open drain output is an error flag output which goes low when V_{OUT} drops 5% below its nominal voltage. This pin also provides a power-on-reset signal if a capacitor is connected to the DELAY pin.
8	$\overline{SD}$	By applying less than 0.4 V to this pin, the device will be turned off. Connect this pin to V_{IN} if unused.
	Exposed Pad	The die substrate is attached to the exposed pad and must be electrically connected to GND.

ORDERING INFORMATION

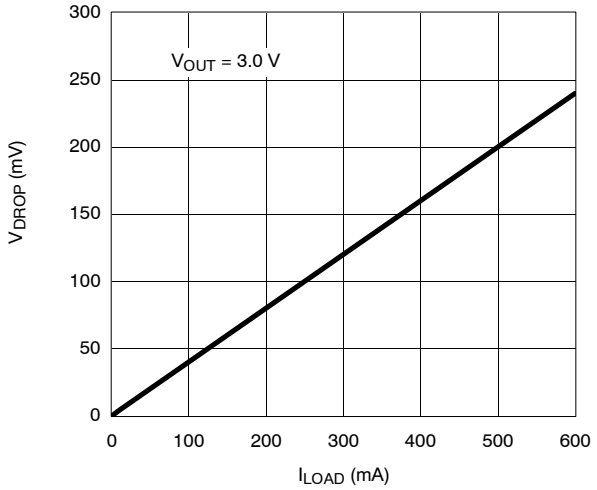
Standard Part Number	Lead (Pb)-Free Part Number	Marking	Voltage	Temperature	Package
Si9185DMP-12-T1	Si9185DMP-12-T1—E3	8512	1.215 V	–40 to 85°C	MLP33 PowerPAK
Si9185DMP-15-T1	Si9185DMP-15-T1—E3	8515	1.50 V		
Si9185DMP-18-T1	Si9185DMP-18-T1—E3	8518	1.80 V		
Si9185DMP-20-T1	Si9185DMP-20-T1—E3	8520	2.00 V		
Si9185DMP-25-T1	Si9185DMP-25-T1—E3	8525	2.50 V		
Si9185DMP-28-T1	Si9185DMP-28-T1—E3	8528	2.80 V		
Si9185DMP-29-T1	Si9185DMP-29-T1—E3	8529	2.90 V		
Si9185DMP-30-T1	Si9185DMP-30-T1—E3	8530	3.00 V		
Si9185DMP-33-T1	Si9185DMP-33-T1—E3	8533	3.30 V		
Si9185DMP-50-T1	Si9185DMP-50-T1—E3	8550	5.00 V		
Si9185DMP-AD-T1	Si9185DMP-AD-T1—E3	85AD	Adjustable		

Additional voltage options are available.

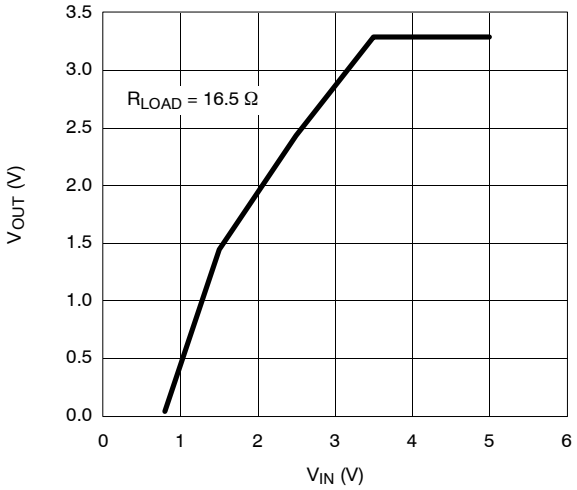
Eval Kit	Temperature Range	Board Type
Si9185DB	–40 to 85°C	Surface Mount

TYPICAL CHARACTERISTICS (INTERNALLY REGULATED, 25°C UNLESS NOTED)

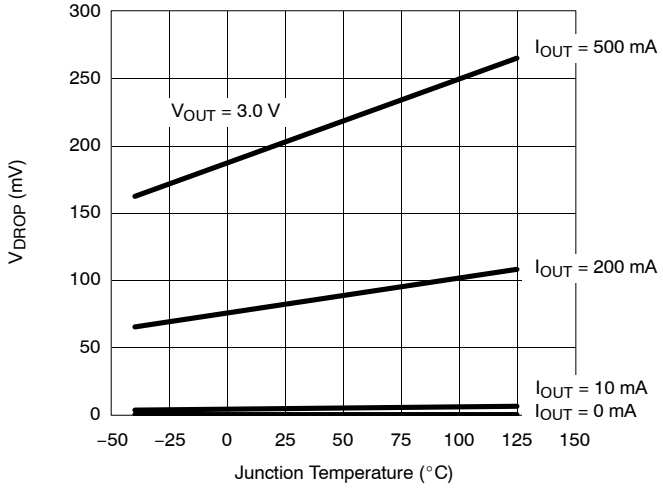
Dropout Voltage vs. Load Current



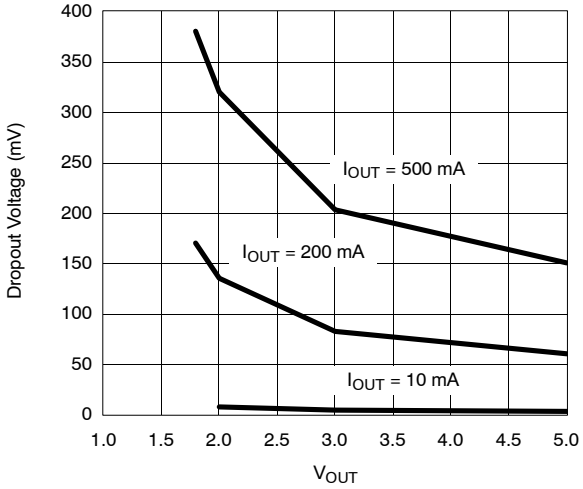
Dropout Characteristic



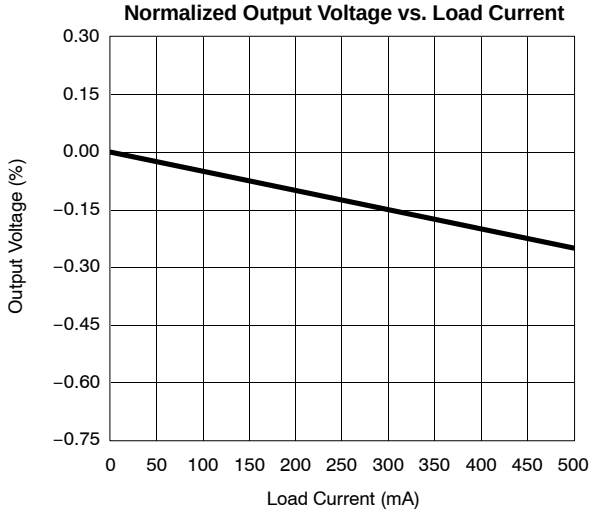
Dropout Voltage vs. Temperature



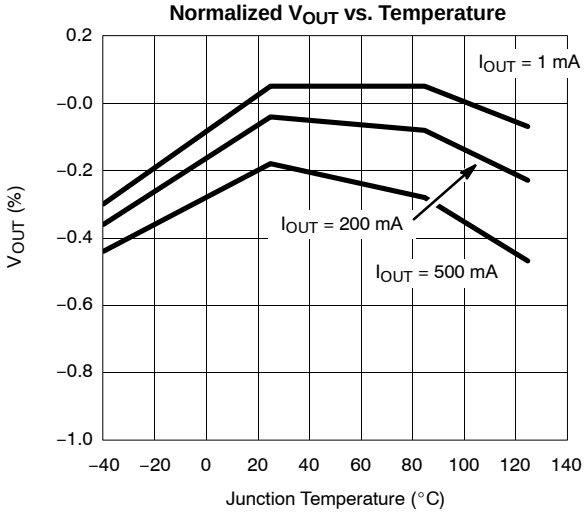
Dropout Voltage vs. V_{OUT}



Normalized Output Voltage vs. Load Current

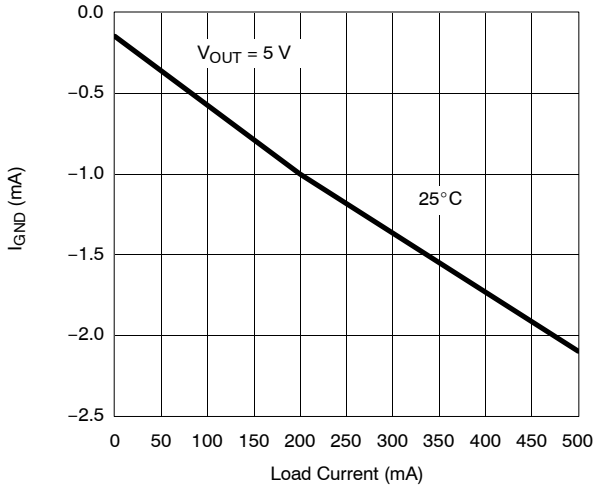


Normalized V_{OUT} vs. Temperature

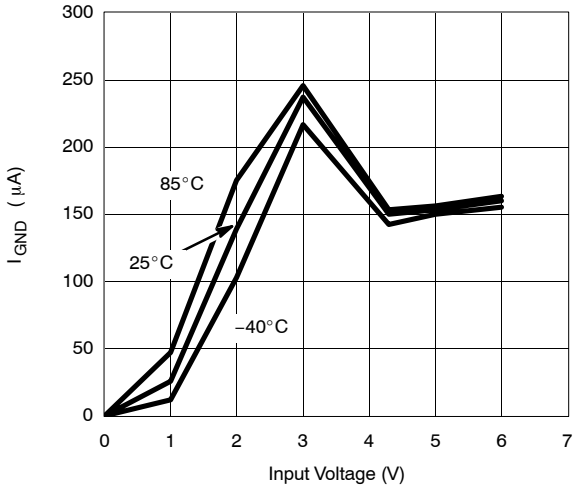


TYPICAL CHARACTERISTICS (INTERNALLY REGULATED, 25°C UNLESS NOTED)

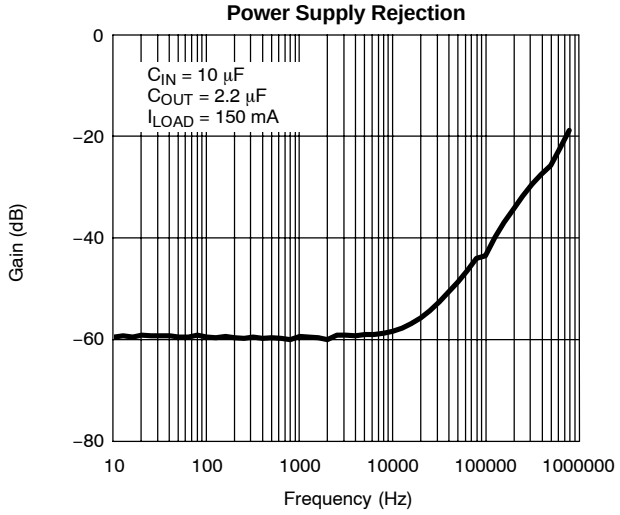
GND Current vs. Load Current



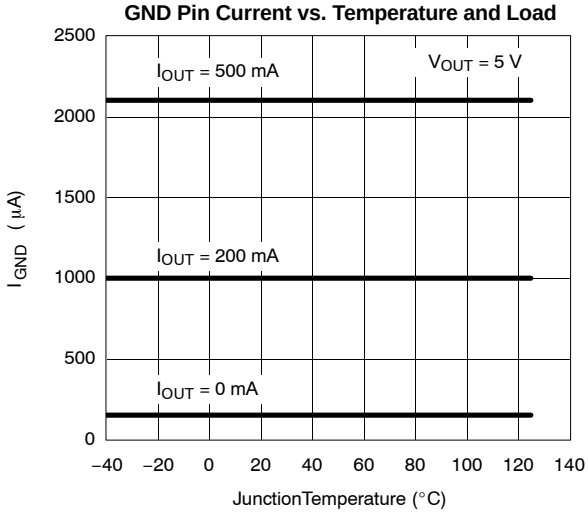
No Load GND Pin Current vs. Input Voltage



Power Supply Rejection

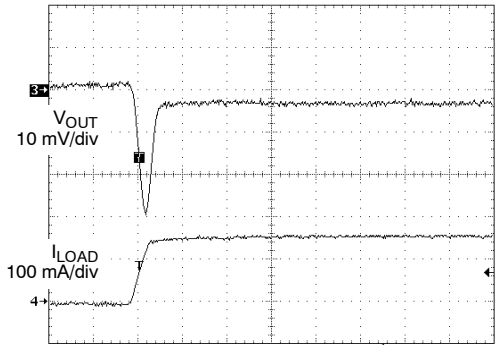


GND Pin Current vs. Temperature and Load



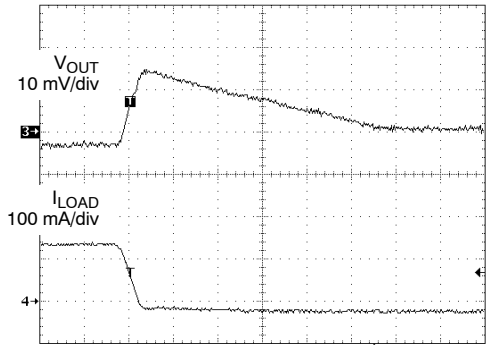
TYPICAL WAVEFORMS

Load Transient Response-1



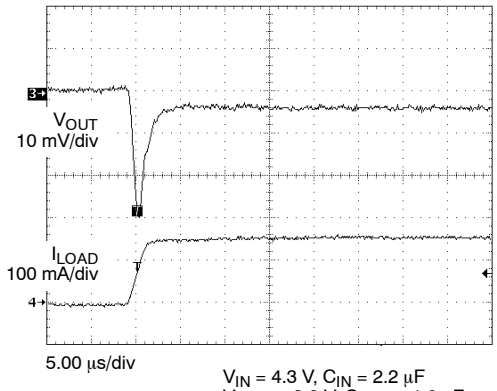
V_{IN} = 4.3 V, C_{IN} = 2.2 µF
V_{OUT} = 3.3 V, C_{OUT} = 2.2 µF
I_{LOAD} = 1 to 150 mA
t_{rise} = 2 µsec

Load Transient Response-2



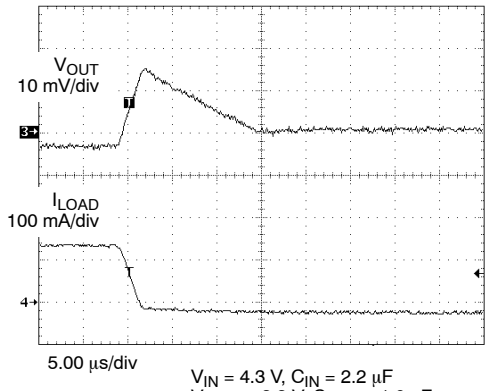
V_{IN} = 4.3 V, C_{IN} = 2.2 µF
V_{OUT} = 3.3 V, C_{OUT} = 2.2 µF
I_{LOAD} = 1 to 150 mA
t_{rise} = 2 µsec

Load Transient Response-3



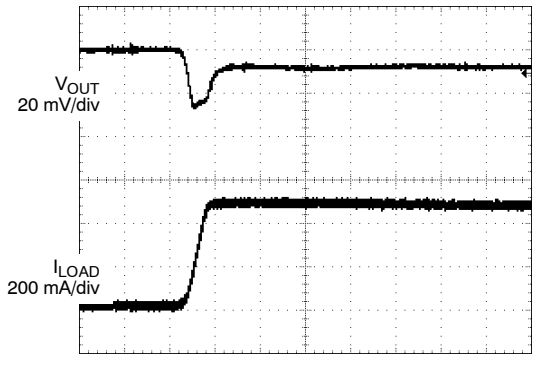
V_{IN} = 4.3 V, C_{IN} = 2.2 µF
V_{OUT} = 3.3 V, C_{OUT} = 1.0 µF
I_{LOAD} = 1 to 150 mA
t_{rise} = 2 µsec

Load Transient Response-4



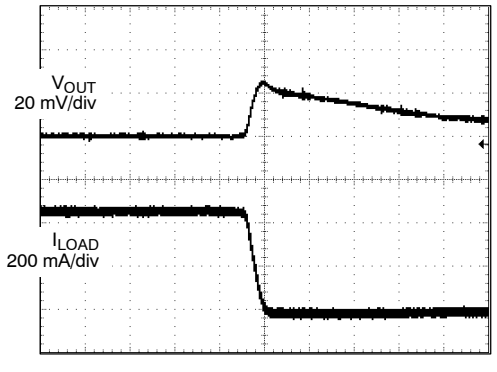
V_{IN} = 4.3 V, C_{IN} = 2.2 µF
V_{OUT} = 3.3 V, C_{OUT} = 1.0 µF
I_{LOAD} = 1 to 150 mA
t_{rise} = 2 µsec

Load Transient Response-5



V_{IN} = 4.3 V, C_{IN} = 10 µF
V_{OUT} = 3.3 V, C_{OUT} = 10 µF
I_{LOAD} = 1 to 500 mA
t_{rise} = 2 µsec

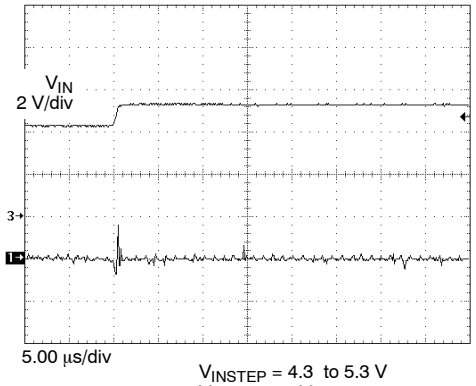
Load Transient Respons-6



V_{IN} = 4.3 V, C_{IN} = 10 µF
V_{OUT} = 3.3 V, C_{OUT} = 10 µF
I_{LOAD} = 1 to 500 mA
t_{rise} = 2 µsec

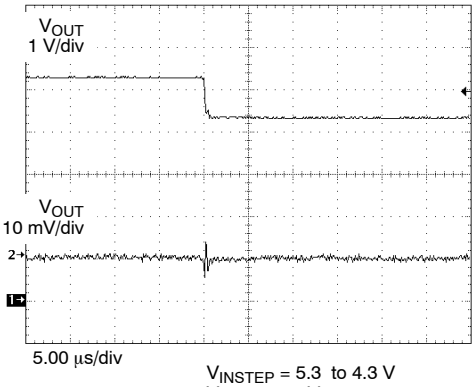
TYPICAL WAVEFORMS

Line Transient Response-1



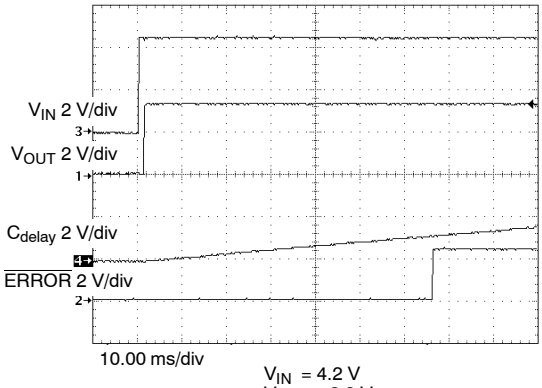
V_{INSTEP} = 4.3 to 5.3 V
V_{OUT} = 3.3 V
C_{OUT} = 2.2 µF
C_{IN} = 10 µF
I_{LOAD} = 500 mA
t_{rise} = 5 µsec

Line Transient Respons-2



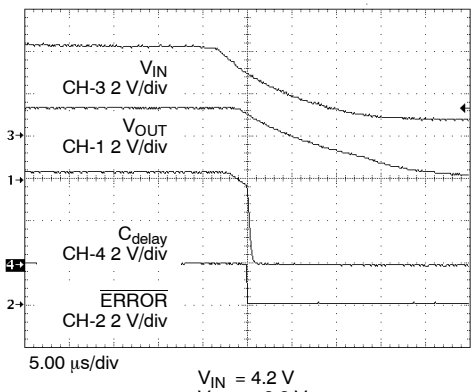
V_{INSTEP} = 5.3 to 4.3 V
V_{OUT} = 3.3 V
C_{OUT} = 2.2 µF
C_{IN} = 10 µF
I_{LOAD} = 500 mA
t_{fall} = 5 µsec

Turn-On Sequence



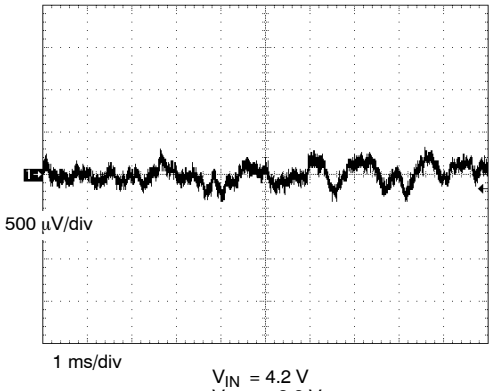
V_{IN} = 4.2 V
V_{OUT} = 3.3 V
C_{delay} = 0.1 µF
C_{NOISE} = 0.1 µF
I_{LOAD} = 350 mA

Turn-Off Sequence



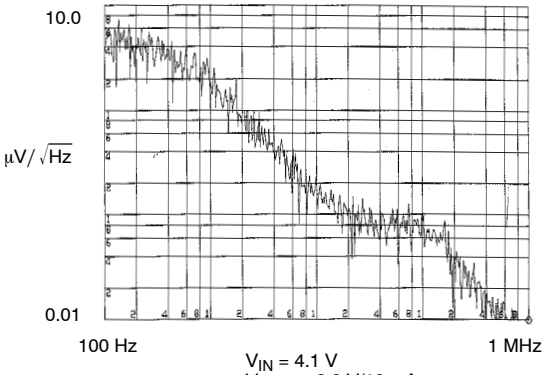
V_{IN} = 4.2 V
V_{OUT} = 3.3 V
C_{delay} = 0.1 µF
C_{NOISE} = 0.1 µF
I_{LOAD} = 350 mA

Output Noise



V_{IN} = 4.2 V
V_{OUT} = 3.3 V
I_{OUT} = 150 mA
C_{NOISE} = 0.1 µF
BW = 10 Hz to 1 MHz

Noise Spectrum



V_{IN} = 4.1 V
V_{OUT} = 3.3 V/10 mA
C_{NOISE} = 0.1 µF

BLOCK DIAGRAM

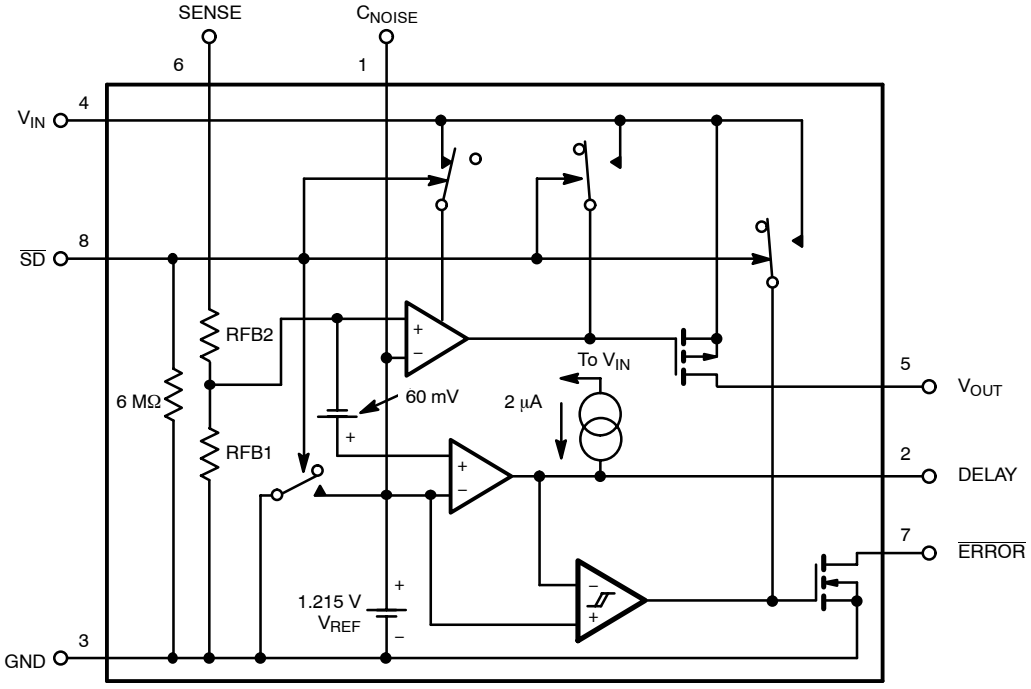


FIGURE 5.

DETAILED DESCRIPTION

The Si9185 is a low drop out, low quiescent current, and very linear regulator family with very fast transient response. It is primarily designed for battery powered applications where battery run time is at a premium. The low quiescent current allows extended standby time while low drop out voltage enables the system to fully utilize battery power before recharge. The Si9185 is a very fast regulator with bandwidth exceeding 50 kHz while maintaining low quiescent current at light load conditions. With this bandwidth, the Si9185 is the fastest LDO available today. The Si9185 is stable with any output capacitor type from 1 μF to 10.0 μF. However, X5R or X7R ceramic capacitors are recommended for best output noise and transient performance.

V_{IN}

V_{IN} is the input supply pin. The bypass capacitor for this pin is not critical as long as the input supply has low enough source impedance. For practical circuits, a 1.0-μF or larger ceramic capacitor is recommended. When the source impedance is not low enough and/or the source is several inches from the Si9185, then a larger input bypass capacitor is needed. It is required that the equivalent impedance (source impedance, wire, and trace impedance in parallel with input bypass capacitor impedance) must be smaller than the input impedance of the Si9185 for stable operation. When the source impedance, wire, and trace impedance are unknown, it is recommended that an input bypass capacitor be used of a value that is equal to or greater than the output capacitor.

V_{OUT}

V_{OUT} is the output voltage of the regulator. Connect a bypass capacitor from V_{OUT} to ground. The output capacitor can be any value from 1.0 μF to 10.0 μF. A ceramic capacitor with X5R or X7R dielectric type is recommended for best output noise, line transient, and load transient performance.

GND

Ground is the common ground connection for V_{IN} and V_{OUT}. It is also the local ground connection for C_{NOISE}, DELAY, SENSE or ADJ, and $\overline{SD}$.

SENSE or ADJ

SENSE is used to sense the output voltage. Connect SENSE to V_{OUT} for the fixed voltage version. For the adjustable output version, use a resistor divider R1 and R2, connect R1 from V_{OUT} to ADJ and R2 from ADJ to ground. R2 should be in the 25-kΩ to 150-kΩ range for low power consumption, while maintaining adequate noise immunity.

The formula below calculates the value of R1, given the desired output voltage and the R2 value,

$$R1 = \frac{(V_{OUT} - V_{ADJ})R2}{V_{ADJ}} \quad (1)$$

V_{ADJ} is nominally 1.215 V.

SHUTDOWN ($\overline{SD}$)

$\overline{SD}$ controls the turning on and off of the Si9185. V_{OUT} is guaranteed to be on when the $\overline{SD}$ pin voltage equals or is greater than 1.5 V. V_{OUT} is guaranteed to be off when the $\overline{SD}$ pin voltage equals or is less than 0.4 V. During shutdown mode, the Si9185 will draw less than 2-µA current from the source. To automatically turn on V_{OUT} whenever the input is applied, tie the $\overline{SD}$ pin to V_{IN}.

$\overline{ERROR}$

$\overline{ERROR}$ is an open drain output that goes low when V_{OUT} is less than 5% of its normal value. As with any open drain output, an external pull up resistor is needed. When a capacitor is connected from DELAY to GROUND, the error signal transition from low to high is delayed (see Delay section). This delayed error signal can be used as the power-on reset signal for the application system. (Refer to Figure 4.)

The $\overline{ERROR}$ pin is disconnected if not used.

DELAY

A capacitor from DELAY to GROUND sets the time delay for $\overline{ERROR}$ going from low to high state. The time delay can be calculated using the following formula:

$$T_{delay} = \frac{(V_{ADJ})C_{delay}}{I_{delay}} \quad (2)$$

The DELAY pin should be an open circuit if not used.

C_{NOISE}

For low noise application, connect a high frequency ceramic capacitor from C_{NOISE} to ground. A 0.01-µF or a 0.1-µF X5R or X7R is recommended.

Safe Operating Area

The ability of the Si9185 to supply current is ultimately dependent on the junction temperature of the pass device. Junction temperature is in turn dependent on power dissipation in the pass device, the thermal resistance of the package and the circuit board, and the ambient temperature. The power dissipation is defined as

$$P_D = (V_{IN} - V_{OUT}) * I_{OUT}.$$

Junction temperature is defined as

$$T_J = T_A + ((P_D * (R\theta_{JC} + R\theta_{CA})).$$

To calculate the limits of performance, these equations must be rewritten.

Allowable power dissipation is calculated using the equation

$$P_D = (T_J - T_A)/ (R\theta_{JC} + R\theta_{CA})$$

While allowable output current is calculated using the equation

$$I_{OUT} = (T_J - T_A)/ (R\theta_{JC} + R\theta_{CA}) * (V_{IN} - V_{OUT}).$$

Ratings of the Si9185 that must be observed are

T_{Jmax} = 125 °C, T_{Amax} = 85 °C, $(V_{IN} - V_{OUT})_{max}$ = 5.3 V, $R\theta_{JC}$ = 4 °C/W.

The value of $R\theta_{CA}$ is dependent on the PC board used. The value of $R\theta_{CA}$ for the board used in device characterization is approximately 46 °C/W.

Figure 6 shows the performance limits graphically for the Si9185 mounted on the circuit board used for thermal characterization.

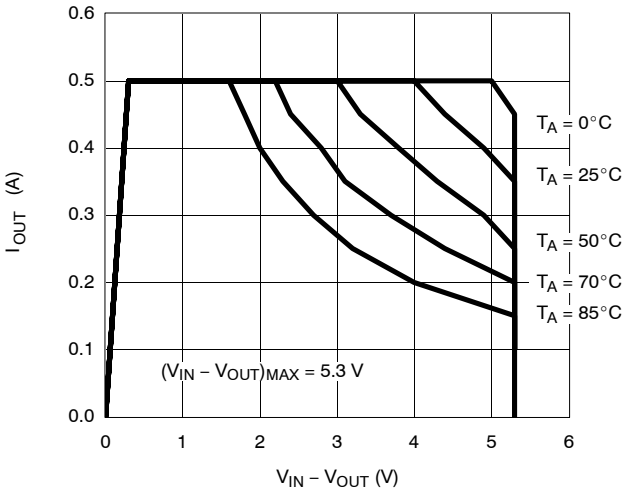


Figure 6.

PCB Footprint and Layout Considerations

The Si9185 comes in the MLP33 PowerPAK package with an exposed pad on the bottom to provide a low thermal impedance path into the PC board. When the PC board layout is designed, a copper plane, referred to as spreading copper, is recommended to be placed under the package to which the exposed pad is soldered. This spreading copper is the path for the heat to move away from the package into the PC board. With the Si9185 mounted on a four layer board measuring 2" × 2", a spreading copper area of 0.25 square inches will yield an $R\theta_{ja}$ of 50°C/W. This allows for power dissipation in excess of 1 watt in an 80°C ambient environment.

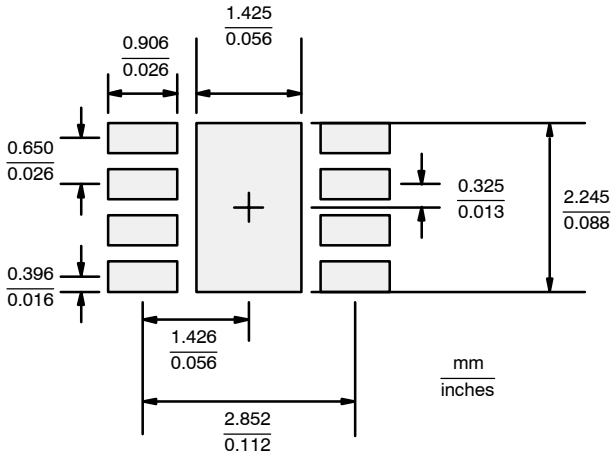


Figure 7. MLP33 PowerPAK Pad Pattern

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