

#### **Features**

- Max. 11,000 counts resolution
- Conversion rate selectable by MPU command: 1.6/s → 128/s
- Input signal full scale: 110mV
- 50/60Hz line noise rejection selectable by MPU command
- Low battery detection
- 3-wire serial bus and EOC signal for MPU I/O port
- -3V power operation with internal charge pumping circuit
- MPU I/O power level selectable by external control pin
- Zero calibration for eliminating offset error
- On-chip buzzer driving and frequency selectable by MPU command
- Support sleep mode by external CS(chipselect) pin

### **Description**

ES51991 is an 11000-count dual-slope analog-to-digital converter (ADC). The conversion rate and buzzer frequency can be selected or decided by an external microprocessor. The conversion rate can be varied from 1.6 time/sec to 128 times/sec under 4MHz/12MHz crystal oscillation clock. Besides, ES51991 also provides low battery detection, power-down mode, 50/60Hz line noise rejection selection, and I/O port level selection for flexible design.

# **Application**

Clamp meter

Thermometer

Portable instrumentation

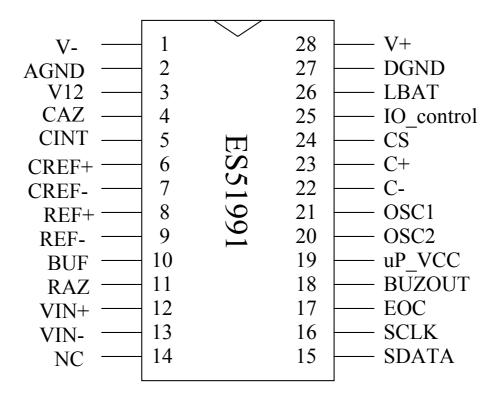
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# **Pin Assignment**

SSOP-28L



# **Pin Description**

Pin No	Symbol	Type	Description		
1	V-	P	Negative supply voltage. Connecting to 3V battery negative terminal.		
2	AGND	G	Analog ground		
3	V12	О	Output of band-gap voltage reference. Typically -1.23V		
4	CAZ	I	Auto-zero capacitor connection.		
5	CINT	О	Integrator output. Connect to integral capacitor		
6	CREF+	I/O	Positive connection for reference capacitor.		
7	CREF-	I/O	Negative connection for reference capacitor.		
8	REF+	I	Differential reference high voltage input.		
9	REF-	I	Differential reference low voltage input.		
10	BUF	О	Buffer output pin. Connect to integral resistor		
11	RAZ	О	Buffer output pin in high-speed mode. Connect to high-speed integral		
			resistor.		
12	VIN+	I	Analog differential high signal input.		
13	VIN-	I	Analog differential low signal input.		
14	NC	-	Not connected		
15	SDATA	I/O	Serial data I/O pin. Nch open-drain output.		
16	SCLK	I	Serial clock input pin.		
17	EOC	О	An indicator for ADC conversion ending.		
18	BUZout	О	Buzzer frequency output		
19	uP_VCC	I	MPU I/O port power level selection		
20	OSC2	О	Crystal oscillation connection		
21	OSC1	I	Crystal oscillation connection		
22	C-	О	Negative capacitor connection for on-chip DC-DC converter.		
23	C+	О	Positive capacitor connection for on-chip DC-DC converter.		
24	CS	I	Chip select input pin. Pull to Low to enter power down mode.		
25	I/O_control	I	MPU I/O port ground level selection		
26	LBAT	I	Low battery configuration. If 3V battery is used, connect it to AGND.		
			The default low-battery threshold voltage is –2.3V. If 9V or other		
			battery voltage is used, the low battery annunciator is displayed when		
			the voltage of this pin is less than V12		
27	DGND	G	Digital ground		
28	V+	O/P	Output of on-chip DC-DC converter.		



#### **Function description**

#### 1. Dual Slope A/D—Four Phases Timing

ES51991 is a dual-slope analog-to-digital converter (ADC). Figure 1 is a structure of dual-slope integrator. Its measurement cycle has two distinct phases: input signal integration (INT) phase and reference voltage integration (DINT) phase.

In INT phase, the input signal is integrated for a fixed time period, then A/D enters DINT phase in which an opposite polarity constant reference voltage is integrated until the integrator output voltage becomes to zero. Since both the time period for input signal integration and the amount of reference voltage are fixed, thus the de-integration time is proportional to the input signal. Hence, we can define the mathematical equation about input signal, reference voltage integration (see Figure 1.):

$$\frac{1}{Buf \times C \text{ int}} \int_{0}^{T_{INT}} V_{IN}(t) dt = \frac{1}{Buf \times C \text{ int}} \times V_{REF} \times T_{DINT}$$

where,  $V_{IN}(t)$  = input signal

 $V_{REF}$  = reference voltage

 $T_{INT}$  = integration time (fixed)

 $T_{\mathit{DINT}}$  = de-integration time (proportional to  $V_{\mathit{IN}}(t)$  )

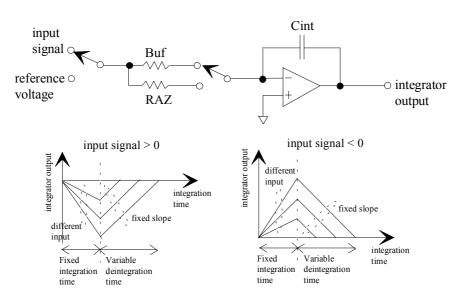


Figure 1. the structure of dual-slope integrator and its output waveform.

If  $V_{IN}(t)$  is a constant, we can rewrite above equation:

$$T_{\mathit{DINT}} = \frac{T_{\mathit{INT}}}{V_{\mathit{REF}}} \times V_{\mathit{IN}}$$

Besides the INT phase and DINT phase, ES51991 exploits auto zero (AZ) phase and zero integration (ZI) phase to achieve accurate measurement. In AZ phase, the system offset is stored. The offset error will be eliminated in DINT phase. Thus a higher accuracy could be obtained. In ZI phase, the internal status will be recovered quickly to that of zero input. Thus the succeeding measurements won't be disturbed by current measurement especially in case of overload.

As mentioned above, the measurement cycle of ES51991 contains four phases:

- (1) auto zero phase (AZ)
- (2) input signal integration phase (INT)
- (3) reference voltage integration phase (DINT)
- (4) zero integration phase (ZI)

The time ratios of these four phases, AZ, INT, DINT and ZI to the entire measurement cycle are 8.8%, 32%, 35.2% and 24% respectively. However the actual duration of each phase depends on conversion rate. An example is shown in the table below. A user can easily deduce other cases based on the table.

#### Voltage:

CR (times/sec)	ZI (ms)	AZ (ms)	INT (ms)	DINT (ms)
8	30	11	40	44

Note: reference voltage = -100 mV.

#### Voltge+PEAK:

CR (times/sec)	ZI (ms)	AZ (ms)	INT (ms)	DINT (ms)
8	30	11	40	60

Note: reference voltage = -100 mV.

#### 2. Component Value Selection for ADC

For various application requirements on conversion rate and input full range, we suggest nominal values for external components of ADC in Figure 1 to obtain better performance. Under default condition with operating clock = 12 MHz:

- (1) conversion rate = 8 times/sec
- (2) reference voltage = -100 mV
- (3) input signal full scale = 110 mV (sensitivity = 10 uV)

We suggest that Cint = 68 nF, Buf = 56 k $\Omega$ 

If a user selects a different conversion rate rather than default, the integration capacitor Cint value must be changed according to the following rule for better performance:

Cint  $\times$  (conversion rate) = (68 nF)  $\times$  (8 times/sec).

A smaller Cint reduces the input full range. However a larger Cint might have weaker noise immunity than the suggested one.

A user could enlarge the input full range by changing reference voltage (Vref) and the amount of integration resistor (Buf). For example, if Vref & Buf are enlarged as twice than the default values then the input full range becomes 220 mV. The input full range can be enlarged up to 1.1V (10 times than the default case). We list general rules in below which might be helpful in determining component values.

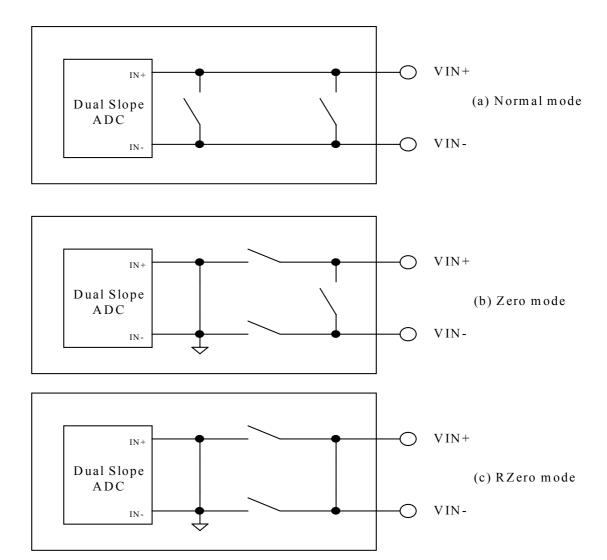
Buf / (reference voltage) =  $56 \text{ k}\Omega$  / (-100 mV)



# 3. Special function

#### 3.1 Zero and RZero Calibration

The Zero and RZero calibration are designed for removing the error rise from the propagation delay of internal component. In Zero or RZero calibration mode, ES51991 outputs a calibration value. The normal measurement value must minus the calibration value to cancel the error and obtain a more accurate value. The following block diagram performs the difference between basic structures of normal mode, Zero calibration and RZero Calibration. We suggest users to do zero-calibration in most applications.





### 3.2 Buzzer Setup

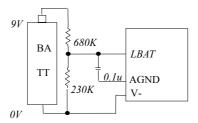
When the bit **BUZ** of ID Byte is set to "H", the BUZOUT will output a square signal of MPU I/O swing level to drive a external buzzer. The buzzer frequency is determined by the bits B0/B1/B2 of STATUS Byte3. The configuration of buzzer frequency is listed at the following table.

B2/B1/B0	BUZout (kHz)
111	4.00
110	3.33
101	3.08
100	2.67
011	2.22
010	2.00
001	1.33
000	1.00

### 3.3 Low Battery Detection

In a case of 3V battery power, the pin LBAT must be shorted to AGND. And the system will have low battery detection level about 2.3V. In another case of 9V or other battery power, the low battery detection happens when the voltage of LBAT is less than –1.23V below GND. And the bit LBAT of STATUS Byte3 will be set to high. A recommended application is shown as following:

### Low battery test (9V)



The low battery detection level is around 7V

#### 3.4 Sleep Mode

When the pin CS is connected to V- or GND (depended on I/O\_control level), the ES51991 will enter sleep mode. In Sleep mode, the chip draws a little supply current. It could extend the battery life. To leave sleep mode or stay in normal mode, the pin CS must be connected to AGND or floating.

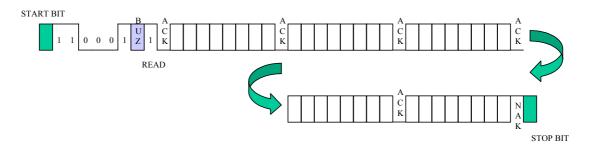
# 4. MPU I/O function definition

#### Write command:

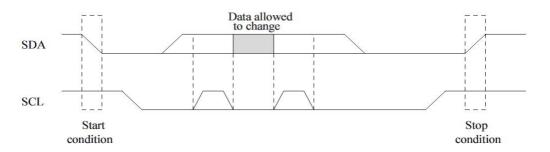
ID byte, Status byte1, Status byte2, Status byte3

#### Read command:

ID byte, Status byte1, Status byte2, Status byte3, Data byte1, Data byte2



# Start and Stop bit



#### ID byte:

1	1	0	0	0	1	BUZ	R/W	
Status by	te1:							
0	0	C0	<b>C</b> 1	<b>C2</b>	SIGN	SEL4M	X	
Status by	Status byte2:							
S60	<b>RZERO</b>	ZERO	0	X	X	X	X	
Status by	te3:							
0	0	<b>B0</b>	<b>B</b> 1	<b>B2</b>	LBAT	X	X	
Data byte	:1:							
<b>D0</b>	<b>D1</b>	<b>D2</b>	<b>D3</b>	D4	<b>D5</b>	<b>D6</b>	<b>D7</b>	
Data byte	2:							
D8	<b>D9</b>	D10	D11	D12	D13	X	X	

**R/W**: set to "H" is in read mode, set to "L" is in write mode

C2/C1/C0/S60: Conversion rate selection, the default is [0000]

C2/C1/C0	S60			
C2/C1/C0	L	Н		
101	128/s	128/s		
100	96/s	96/s		
011	64/s	76.8/s		
010	32/s	38.4/s		
001	16/s!	19.2/s*		
000	8/s!	9.6/s*		
110	3.2/s <sup>!*</sup>	3.84/s*		
111	1.6/s!*	1.92/s*		

Crystal: 12MHz

!: 50Hz line noise rejection, \*: 60Hz line noise rejection

SEL4M: "H" is XTAL is 4MHz version, "L" is default 12MHz XTAL

C2/C1/C0	S60
C2/C1/C0	X
101	128/s
100	64/s
011	64/s
010	32/s
001	16/s!
000	8/s!
110	3.2/s <sup>!*</sup>
111	1.6/s <sup>!*</sup>

Crystal:4MHz

**SIGN**: "H" is negative, "L" is positive

**LBAT**: "H" is low battery detection flag active, the default is "L"

**RZERO**: "H" is RZero calibration mode "ON", the default is "L"

**ZERO**: "H" is Zero calibration mode "ON", the default is "L"

**B2/B1/B0**: Buzzer frequency selection (independent with conversion rate)

**BUZ**: "H" is buzzer turn on and "L" is turn off, the default is turn off.

#### Buzzer ON

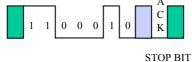
START BIT

1 1 0 0 0 1 1 C K

STOP BIT

#### **Buzzer OFF**

START BIT



**D13-D0**: ADC output data. Binary code format.

### 5. Power and I/O output level selection

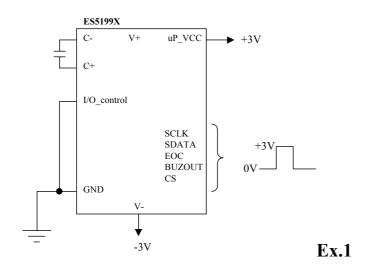
#### Power

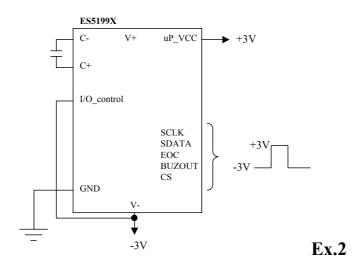
- Charge pump output for positive supply voltage(V+)
- External DC source to V+ is available by floating the charge pump capacitor

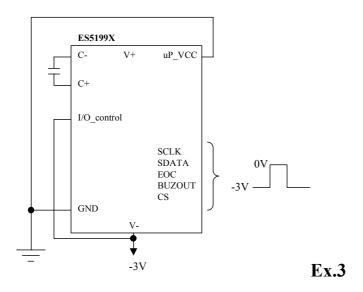
#### I/O output level selectable

- uP\_VCC provided by external DC source (the same high level with MPU)
- A control pin (I/O control) selects the low level to -3V(V-) or 0V(DGND)

nD VCC	I/O control	I/O level		Evample
uP_VCC	I/O_control	Н	L	Example
3	Н	+3V	0V	Ex.1
3	L	+3V	-3V	Ex.2
0	L	0V	-3V	Ex.3









# **Absolute Maximum Ratings**

Characteristic	Rating
Supply Voltage (V- to AGND)	-3.6V
Analog Input Voltage	V0.6 to $V + +0.6$
V+	$V+ \ge (AGND/DGND+0.5V)$
AGND/DGND	$AGND/DGND \ge (V0.5V)$
Digital Input	V0.6 to DGND +0.6 or V+ +0.6
Power Dissipation. Flat Package	500mW
Operating Temperature	$0^{\circ}$ C to $70^{\circ}$ C
Storage Temperature	-25°C to 125°C

### **DC Electrical Characteristics**

 $TA=25^{\circ}C$ ,  $V_{CM}=0V$ ,  $V_{-}=-3V$ 

Parameter	Symbol	Test Condition	Min.	Тур.	Max	Units
Power supply	V-		-3.3	-3.0	-2.5	V
Operating supply current Conversion rate = 8/sec.	$I_{DD}$	Normal operation (XTAL=12MHz)		2.0	2.2	mA
	$I_{SS}$	In sleep mode		2.5	5	μΑ
Voltage roll-over error	REV		_	_	±0.05	%F.S <sup>1</sup>
Voltage nonlinearity	NLV	Best case straight line	_	_	±0.05	%F.S
Input Leakage				1	10	pA
Low battery flag voltage		V- to AGND	-2.4	-2.3	-2.2	V
Internal pull high to up. Was current		CS (uP_Vcc=3V)		5		11 A
Internal pull-high to uP_Vcc current		CS(uP_Vcc=0V)		1.5		uA
Internal pull-low to V- current		I/O_control (V-=-3V)		1.5		uA
Zero input reading		$10M\Omega$ input resistor zero cal. by MPU	-000	000	+000	counts
Reference voltage and open circuit voltage for $110\Omega$ measurement	$ m V_{REF}$	$100 K\Omega$ resistor between VRH and AGND	-1.33	-1.23	-1.13	V
Reference voltage temperature coefficient	$TC_{RF}$	$100$ KΩ resister Between VRH $0^{\circ}$ C < TA < $70^{\circ}$ C	_	50	_	ppm/°C

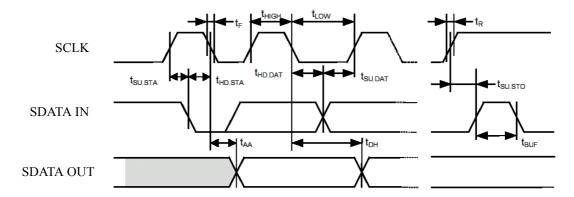
Note:

1.Full Scale

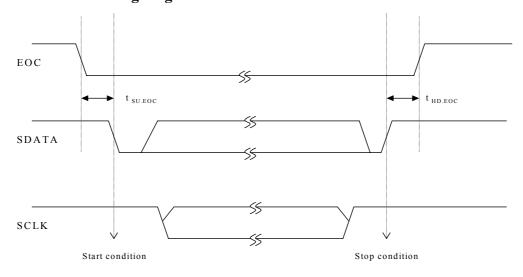
# **AC Electrical Characteristics**

Parameter	Symbol	Min.	Тур.	Max.	Unit
SCLK clock frequency	$f_{SCLK}$	-	-	100	kHz
SCLK clock time "L"	$t_{LOW}$	4.7	ı	-	
SLCK clock time "H"	$t_{ ext{HIGH}}$	4.0	1	-	us
SDATA output delay time	$t_{AA}$	0.1	1	3.5	
SDATA output hold time	$t_{ m DH}$	100	1	-	ns
Start condition setup time	$t_{SU.STA}$	4.7	1	-	110
Start condition hold time	$t_{ m HD.STA}$	4.0	1	- us	
Data input setup time	$t_{SU.DAT}$	200	1	-	nc
Data input hold time	$t_{ m HD.DAT}$	0	1	-	ns
Stop condition setup time	$t_{ m SU.STO}$	4.7	1	-	
SCLK/SDATA rising time	$t_{\rm R}$	-	-	1.0	110
SCLK/SDATA falling time	$t_{\scriptscriptstyle \mathrm{F}}$	-	-	0.3	us
Bus release time	$t_{ m BUF}$	4.7	-		
EOC setup time in read mode	$t_{SU.EOC}$	0			ns
EOC hold time in read mode	$t_{\rm HD.EOC}$	0	-	-	ns

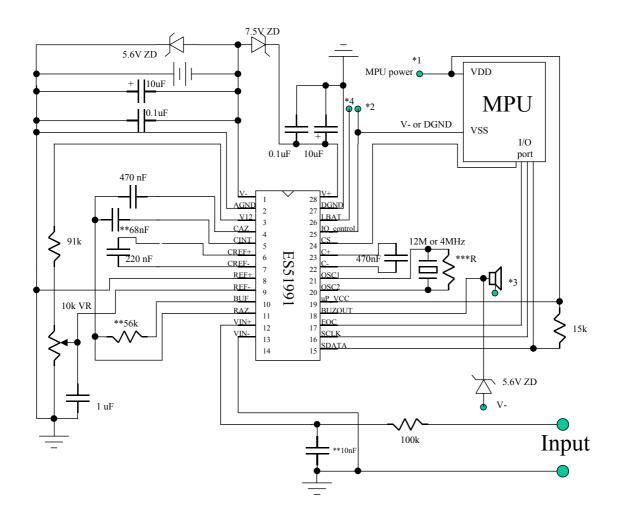
### I/O timing diagram



### Read mode EOC timing diagram



# **Application example**



#### Note:

Zener diodes in above circuit are used for IC protection, so MUST be soldered on PCB first.

\*1\*2\*3\*4: Depend on power design

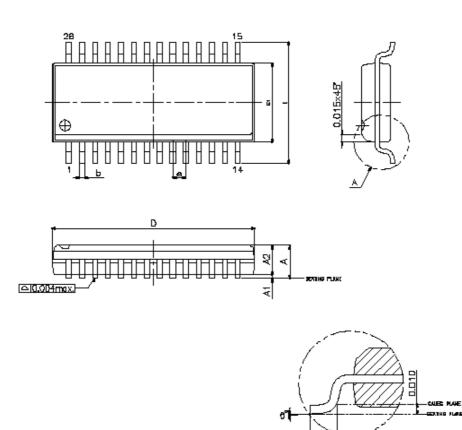
\*\* Depends on conversion rates setting: V-=-3.0V

(a)Conversion rate	$(b)C_{INT}(uF)$	$(c)R_{BUF}(k\Omega)$	(a)Conversion rate	$(b)C_{INT}(uF)$	$(c)R_{BUF}(k\Omega)$
128/s	0.01	22	16/s	0.068	27
96/s	0.01	30	9.6/s	0.047	68
76.8/s	0.01	39	8/s	0.068	56
64/s	0.022	22	3.84/s	0.1	82
38.4/s	0.022	36	3.2/s	0.1	91
32/s	0.033	27	1.92/s	0.22	68
19.2/s	0.033	47	1.6/s	0.22	91

\*\*\*  $R=10\sim22M\Omega$  resistor is optional



**Product Outline: SSOP-28L** 



SYMBOLS	MIN.	MAX.
Α	0.053	0.069
A1	0.004	0.010
A2	ı	0.059
Ь	0.008	0.012
D	0.386	0.394
E1	0.150	0.157
е	0.025 BASIC	
E	0.228	0.244
L	0.016	0.050
а	ā	8

UNIT : NCH

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DETAIL A

NOTES:

1.JEDEC CUITLINE: ND-137 AF

2.DINENSIONS "D" DOES NOT INCLUDE MOLD FLASH,
PROTRUSIONS OR GATE BURRS, MOLD FLASH, PROTRUSIONS
AND GATE BURRS SHALL NOT EXCEED 0.15mm (0.00%)
PER SIDE.

<sup>3.</sup>DNENSIONS "E1" DOES NOT INCLUDE INTER-LEAD FLASH. OR PROTRUSIONS. INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED D.25mm (D.010m) PER SIDE.